# DTMF Generation with a 3.58 MHz Crystal

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DTMF (Dual Tone Multiple Frequency) is associated with digital telephony, and provides two selected output frequencies (one high band, one low band) for a duration of 100 ms. DTMF generation consists of selecting and combining two audio tone frequencies associated with the rows (low band frequency) and columns (high band frequency) of a pushbutton touch tone telephone keypad.

This application note outlines two different methods of DTMF generation using a COP820C/840C microcontroller clocked with a 3.58 MHz crystal in the divide by 10 mode. This yields an instruction cycle time of 2.79  $\mu$ s. The application note also provides a low true row/column decoder for the DTMF keyboard.

The first method of DTMF generation provides two PWM (Pulse Width Modulation) outputs on pins G3 and G2 of the G port for 100 ms. These two PWM outputs represent the selected high band and low band frequencies respectively, and must be combined externally with an LM324 op amp or equivalent feed back circuit to produce the DTMF signal.

The second method of DTMF generation uses ROM lookup tables to simulate the two selected DTMF frequencies. These table lookup values for the selected high band and low band frequencies are then combined arithmetically. The high band frequencies contain a higher bias value to compensate for the DTMF requirement that the high band frequency component be 2 dB above the low band frequency component to compensate for losses in transmission. The resultant value from the arithmetic combination of sine wave values is output on L port pins L0 to L5, and must be combined externally with a six input resistor ladder network to produce the DTMF signal. This resultant value is updated every 118 µs. The COP820C/840C timer is used to time out the 100 ms duration of the DTMF. A timer interrupt at the end of the 100 ms is used to terminate the DTMF output. The external ladder network need not contain any active components, unlike the first method of DTMF generation with the two PWM outputs into the LM324 op amp.

The associated COP820C/840C program for the DTMF generation is organized as three subroutines. The first subroutine (KBRDEC) converts the low true column/row input from the DTMF keyboard into the associated DTMF hexadecimal digit. In turn, this hex digit provides the input for the other two subroutines (DTMFGP and DTMFLP), which represent the two different methods of DTMF generation. These three subroutines contain 35, 94, and 301 bytes of COP820C/840C code respectively, including all associated ROM tables. The Program Code/ROM table breakdowns are 19/16, 78/16, and 88/213 bytes respectively.

#### DTMF KEYBOARD MATRIX

The matrix for selecting the high and low band frequencies associated with each key is shown in *Figure 1*. Each key is uniquely referenced by selecting one of the four low band frequencies associated with the matrix rows, coupled with selecting one of the four high band frequencies associated with the matrix columns. The low band frequencies are 697 Hz, 770 Hz, 852 Hz, and 941 Hz, while the high band frequencies are 1209 Hz, 1336 Hz, 1477 Hz, and 1633 Hz. The DTMF keyboard input decode subroutine assumes that the keyboard is encoded in a low true row/column format, where the keyboard is strobed sequentially with four low true column selects with each returning a low true row select. The low true column and row selects are encoded in the upper and lower nibbles respectively of the accumulator, which serves as the input to the DTMF keyboard input decode subroutine. The subroutine will then generate the DTMF hexadecimal digit associated with the DTMF keyboard input diait.

The DTMF keyboard decode subroutine (KBRDEC) utilizes a common ROM table lookup for each of the two nibbles representing the low true column and row encodings for the keyboard. The only legal low true nibbles for a single key input are E, D, B, and 7. All other low true nibble values represent multiple keys, no key, or no column strobe. Results from two legal nibble table lookups (from the same 16 byte ROM table) are combined to form a hex digit with the binary format of 0000RRCC, where RR represents the four row values and CC represents the four column values. The illegal nibbles are trapped, and the subroutine is exited with a RET (return) command to indicate multiple keys or no key. A pair of legal nibble table lookups result in the subroutine being exited with a RETSK (return and skip) command to indicate a single key input. This KBRDEC subroutine uses 35 bytes of code, consisting of 19 bytes of program code and 16 bytes of ROM table.

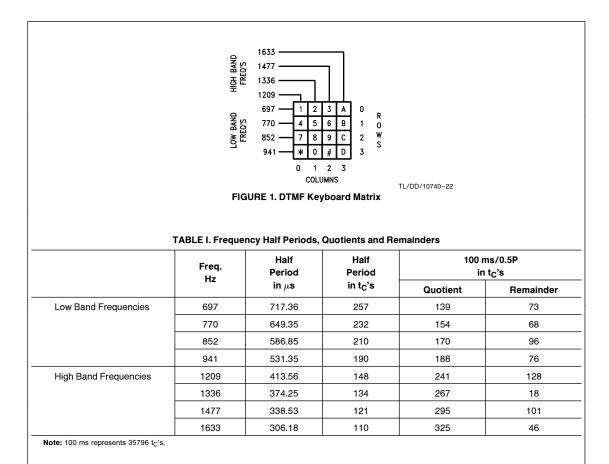
# DTMF GENERATION USING PWM AND AN OP AMP

The first DTMF generation method (using the DTMFGP subroutine) generates the selected high band and low band frequencies as PWM (Pulse Width Modulation) outputs on pins G3 and G2 respectively of the G port. The COP820C/ 840C microcontrollers each contain only one timer, and three times must be generated to satisfy the DTMF application. These three times are the half periods of the two selected frequencies and the 100 ms duration period. Obviously the single timer can only generate one of the required times, while the program must generate the two remaining times. The solution lies in dividing the 100 ms duration time by the half periods for each of the eight DTMF frequencies, and then examining the respective high band and low band quotients and remainders. Naturally these divisions must be normalized to the instruction cycle time (t<sub>C</sub>). 100 ms represents 35796 t<sub>C</sub>'s. The results of these divisions are detailed in Table I

The four high band frequencies are produced by running the COP820C/840C timer in PWM (Pulse Width Modulation) mode, while the program produces the four low band frequencies and the 100 ms duration timeout. The programmed times are achieved by using three programmed register counters R0, R2 and R3, with a backup register R1 to reload the counter R0. These three counters represent the half period, the 100 ms quotient, and the 100 ms remainder associated with each of the four low band frequencies.

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The DTMFGP subroutine starts by transforming the DTMF hex digit in the accumulator (with binary format 0000RRCC) into low and high frequency vectors with binary formats 0011RR11 and 0011CC00 respectively. The transformation of the hex digit 0000RRCC (where RR is the row select and CC is the column select) into the frequency vectors is shown in Table II. The conversion produces a timer vector 0011CC00 (T), and three programmed counter vectors for R1, R2, and R3. The formats for the three counter vectors are 0011RR11 (F), 0011RR10 (Q), and 0011RR01 (R). These four vectors created from the core vector are used as inputs for a 16 byte ROM table using the LAID (Load Accumulator InDirect) instruction. One of these four vectors (the T vector) is a function of the column bits (CC), while the other three vectors (F, Q, R) are a function of the row bits (RR). This correlates to only one parameter being needed for the timer (representing the selected high band frequency), while three parameters are needed for the three counters (half period, 100 ms quotient, 100 ms remainder) associated with the low band frequency and 100 ms duration. The frequency parameter ROM translation table, accessed by the T, F, Q, and R vectors, is shown in Table III.

**R**— Remainder

#### **TABLE II. DTMF Hex Digit Translation**

RCC — — — — — — — — — — — — — — — — — —			-
		*:	*
		*	*
		*	*
		*	*
		*	*
Timer	Т	00110	CC00
R1	F	0011F	R11
R2	Q	0011F	R10
R3	R	0011F	R01
	Timer R1 R2	R1 F R2 Q	* * * Timer T 00110 R1 F 00111 R2 Q 00111

# TABLE III. Frequency Parameter ROM Translation Table

T— Timer	F— Frequency	Q— Quotient	
Address	Data (Decimal)	Vector	
0x30	147	Т	
0x31	10	R	
0x32	140	Q	
0x33	38	F	
0x34	133	Т	
0x35	9	R	
0x36	155	Q	
0x37	33	F	
0x38	120	Т	
0x39	14	R	
0x3A	171	Q	
0x3B	31	F	
0x3C	109	Т	
0x3D	10	R	
0x3E	189	Q	
0x3F	26	F	

The theory of operation in producing the selected low band frequency starts with loading the three counters with values obtained from a ROM table. The half period for the selected frequency is counted out, after which the G2 output bit is toggled. During this half period countout, the quotient counter is decremented. This procedure is repeated until the quotient counter counts out, after which the program branches to the remainder loop. During the remainder loop, the remainder counter counts out to terminate the 100 ms. Following the remainder countout, the G2 and G3 bits are both reset, after which the DTMF subroutine is exited. Great care must be taken in time balancing the half period loop for the selected low band frequency. Furthermore, the toggling of the G2 output bit (achieved with either a set or reset bit instruction) must also be exactly time balanced to maintain the half period time integrity. Local stall loops (consisting of a DRSZ instruction followed by a JP jump back to the DRSZ for a two byte, six instruction cycle loop) are embedded in both the half period and remainder loops. Consequently, the ROM table parameters for the half period and remainder counters are approximately only one-sixth of what otherwise might be expected. The program for the half period loop, along with the detailed time balancing of the loop for each of the low band frequencies, is shown in *Figure 2*.

	Prog		Bytes/ Cycles		ditional rcles	Cycles	Total Cycles
	LD	B, #PORTGD					
	LD	X,#R1	2/3				
LUP1:	LD	A,[X-]	1/3			3	
	IFBIT	2,[B]	1/1			1	
	JP	BYP1	1/3	3	1		
	Х	A,[X+]	1/3		3		
	SBIT	2,[B]	1/1		1		
	JP	BYP2	1/3		3		
BYP1:	NOP		1/1	1			
	RBIT	2,[B]	1/1	1			
	Х	A,[X+]	1/3	3			
BYP2:	DRSZ	R2	1/3			3	
	JP	LUP2	1/3			3	
	JP	FINI	1/3				
LUP2:	DRSZ	R0	1/3		3	3	
	JP	LUP2	1/3		3	1	
	LD	A,[X]	1/3			3	
	IFEQ	A,#31	2/2			2	
	JP	LUP1	1/3		1	3	30
	NOP		1/1		1		
	NOP		1/1		1		
	IFEQ	A,#38	2/2		2		
	JP	LUP1	1/3	1	3		35
	LAID		1/3	3			
	NOP		1/1	1			
	JP	LUP1	1/3	3			40
		Table III	Stall	Total	Half		
			Loop	Cycles	Period		
		[(38 – 1)	× 6]	+ 35	= 257		
		[(33 - 1)	× 6]	+ 40	= 232		
		[(31 - 1)	× 6]	+ 30	= 210		
		[(26 - 1)	× 6]	+ 40	= 190		
		FIGURE 2. Time	-				
		FIGURE 2. TIME	Dalancing	IOI Hall Peri	0u Loop		

Table III	Stall	R Loop	Total	Table I
Remainder	Loop	Overhead	Cycles	Remainder
[(10 - 1)	× 6]	+ 20	= 74	73
[(9-1)	× 6]	+ 20	= 68	68
[(14 - 1)	× 6]	+ 20	= 98	96
[(10 - 1)]	× 6]	+ 20	= 74	76

Note that the Q value in Table III is one greater than the quotient in Table I to compensate for the fact that the quotient count down to zero test is performed early in the half period loop. The overhead in the remainder loop is 20 instruction cycles. The detailed time balancing for the remainder loop is shown in Table IV.

The selected high band frequency is achieved by loading the half period count in t<sub>C</sub>'s minus one (from Table III) into the timer autoreload register and running the timer in PWM output mode. The minus one is necessary since the timer toggles the G3 output bit when it underflows (counts down through zero), at which time the contents of the autoreload register are transferred into the timer.

In summary, the input digit from the keyboard (encoded in low true column/row format) is translated into a digit matrix vector XXXXRRCC which is checked for 1001RRCC to indicate a single key entry. No key or multiple key entries will set a flag and terminate the DTMF subroutine. The digit matrix vector for a single key is transformed into the core vector 0000RRCC. The core vector is then translated into four other vectors (T, F, Q, R) which in turn are used to select four parameters from a 16 byte ROM table. These four parameters are used to load the timer, and the respective half period, quotient, and remainder counters. The 16 byte ROM table must be located starting at ROM location 0030 (or 0X30) in order to minimize program size, and has reference setups with the "OR A, #033" instruction for the F vector and the "OR A, #030" instruction for the T vector.

The three parameters associated with the two R bits of the core vector require a multi-level table lookup capability with the LAID instruction. This is achieved with the following section of code in the DTMF subroutine:

1

LUP

JP

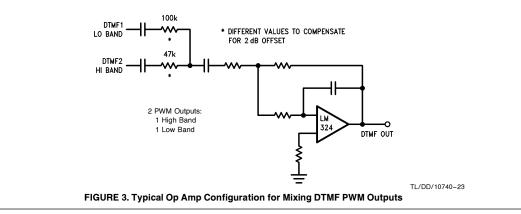
LUP

This program loads the F frequency vector into R1, and then decrements the vector each time around the loop. The vector is successively moved with the exchange commands from R1 to R2 to R3 as one of the same exchange commands loads the data from the ROM table into R1, R2, and R3. This successive decrementation of the F vector changes the F vector into the Q vector, and then changes the Q vector into the R vector. These vectors are used to access the ROM table with the LAID instruction. The B pointer is incremented each time around the loop after it has been used to store away the three selected ROM table parameters (one per loop). These three parameters are stored in sequential RAM locations R1, R2, and R3. The IFBNE test instruction is used to skip out of the loop once the three selected ROM table parameters have been accessed and stored away.

The timer is initialized to a count of 15 so that the first timer underflow and toggling of the G3 output bit (with timer PWM mode and G3 toggle output selected) will occur at the same time as the first toggling of the G2 output bit. The half period counts for the high band frequencies minus one are stored in the timer section of the ROM table. The selected value from this frequency ROM table is stored in the timer autoreload register. The timer is selected for PWM output mode and started with the instruction LD [B], #0B0 where the B pointer is selecting the CNTRL register at memory location 0FF

This first DTMF generation subroutine for the COP820C/ 840C uses 94 bytes of code, consisting of 78 bytes of program code and 16 bytes of ROM table. A program test routine to sequentially call the DTMFGP subroutine for each of the 16 keyboard input digits is supplied with the listing for the DTMF35 program. This test routine uses a 16 byte ROM table to supply the low true encoded column/row keyboard input to the accumulator. An input from the I0 input pin of the I port is used to select which DTMF generation subroutine is to be used. The DTMFGP subroutine is selected with 10 = 0.

A TYPICAL OP AMP CONFIGURATION FOR MIXING THE TWO DTMF PWM OUTPUTS IS SHOWN IN FIGURE 3.



# DTMF GENERATION USING A RESISTOR LADDER NETWORK

The second DTMF generation method (using the DTMFLP subroutine) generates and combines values from two table lookups simulating the two selected sine waves. The high band frequency table values have a higher base line value (16 versus 13) than the low band frequency table values. This higher bias for the high frequency values is necessary to satisfy the DTMF requirement that the high band DTMF frequencies need a value 2 dB greater than the low band DTMF frequencies to compensate for losses in transmission.

The resultant value from arithmetically combining the table lookup low band and high band frequency values is output on pins L0 to L5 of the L port in order to feed into a six input external resistor ladder network. The resultant value is updated every 1171/<sub>3</sub>  $\mu$ s (one cycle of the LUP42 program loop). The LUP42 program loop contains 42 instruction cycles (t<sub>C</sub>'s) of 2.7936511  $\mu$ s each for a total loop time of 1171/<sub>3</sub>  $\mu$ s. The COP820C/840C timer is used to count out the 100 ms DTMF duration time.

An interrupt from the timer terminates the 100 ms DTMF output. Note that the Stack Pointer (SP) must be adjusted following the timer interrupt before returning from the DTMFLP subroutine.

The DTMFLP subroutine starts by quadrupling the value of the DTMF hex digit value in the accumulator, and then adding an offset value to reach the first value in the telephone key table. The telephone key ROM table contains four values associated with each of the 16 DTMF hex keys. These four values represent the low and high frequency table sizes and table starting addresses associated with the pair of frequencies (one low band, one high band) associated with each DTMF key. The FRLUP section of the program loads the four associated telephone key table values from the ROM table into the registers LFTBSZ (Low Freq Table Size), LFTADR (Low Freq Table Address), HFTBSZ (High Freq Table Size), and HFTADR (High Freq Table Address). The program then initializes the timer and autoreload register, starts the timer, and then jumps to LUP42. Note that the timer value in  $t_C$ 's is 100 ms plus one LUP42 time, since the initial DTMF output is not until the end of the LUP42 program

Multiples of the magic number 118  $\mu$ s (approximately) are close approximations to all eight of the DTMF frequencies. The LUP42 program uses 42 instruction cycles (of 2.7936511  $\mu$ s each) to yield a LUP42 time of 1171/<sub>2</sub>  $\mu$ s. The purpose of the LUP42 program is to update the six L port outputs by accessing and then combining the next set of

values from the selected low band and high band sine wave frequency tables in the ROM. The ROM table offset frequency pointers (LFPTR and HFPTR) must increment each time and then wrap around from top to bottom of the two selected ROM tables. The ROM table size parameters (LFTBSZ and HFTBSZ) for the selected frequencies are tested during each LUP42 to determine if the wrap around from ROM table top to bottom is necessary. The wrap around is implemented by clearing the frequency pointer in question. Note that the ROM tables are mapped from a reference of 0 to table size minus one, so that the table size is used in a direct comparison with the frequency offset pointer to test for the need for a wrap around. Also note that the offset pointer incremented value is used during the following LUP42 cycle, while the pre-incremented value of the pointer is used during the current cycle. However, it is the incremented value that is tested versus the table size for the need to wrap around.

After the low band and high band ROM table sine wave frequency values are accessed in each cycle of the LUP42 program, they are added together and then output to pins LO-L5 of the L port. As stated previously, the low band frequency values have a lower bias than the high band frequency values to compensate for the required 2 dB offset. Specifically, the base line and maximum values for the low frequency values are 13 and 26 respectively, while the base line and maximum values for the high frequency values are 16 and 32 respectively. Thus the combined base line value is 29, while the combined maximum value is 58. This gives a range of values on the L port output (LO-L5) from 0 to 58.

The minimum time necessary for the LUP42 update program loop is 36 instruction cycles including the jump back to the start of the loop. Consequently, two LAID instructions are inserted just prior to the jump back instruction at the end of LUP42 to supply the six extra NOP instruction cycles needed to increase the LUP42 instruction cycles from 36 to 42. A three cycle LAID instruction can always be used to simulate three single cycle NOP instructions if the accumulator data is not needed.

Table V shows the multiple LUP42 approximation to the eight DTMF frequencies, including the number of sine wave cycles and data points in the approximation. As an example, three cycles of a sine wave with a total of 19 data points across the three cycles is used to approximate the 1336 Hz DTMF frequency. The 19 cycles of LUP42 times the LUP42 time of 1171/<sub>3</sub>  $\mu$ s is divided into the three cycles to yield a value of 1345.69 Hz. This gives an error of +0.73% when compared with the DTMF value of 1336 Hz. This is well within the 1.5% North American DTMF error range.

#### **TABLE V. DTMF Frequency Approximation Table**

DTMF Freq.	# of Sine Wave Cycles	# of Data Points	Calculation	Approx. Freq.	% Erro
697	4	49	4/(49 x 1171/3)	= 695.73	-0.18
770	1	11	1/(11 x 1171⁄3)	= 774.79	+ 0.62
852	1	10	1/(10 x 1171/3)	= 852.27	+0.03
941	1	9	1/(9 x 1171/3)	= 946.97	+0.63
1209	1	7	1/(7 x 1171/3)	= 1217.53	+0.71
1336	3	19	3/(19 x 1171/3)	= 1345.69	+0.73
1477	4	23	4/(23 x 1171/3)	= 1482.21	+ 0.35
1633	4	21	4/(21 x 117 <sup>1</sup> / <sub>3</sub> )	= 1623.38	-0.59

The frequency approximation is equal to the number of cycles of sine wave divided by the time in the total number of LUP42 cycles before the ROM table repeats.

The values in the DTMF sine wave ROM tables are calculated by computing the sine value at the appropriate points, scaling the sine value up to the base line value, and then adding the result to the base line value. The following example will help to clarify this calculation.

Consider the three cycles of sine wave across 19 data points for the 1336 Hz high band frequency. The first value in the table is the base line value of 16. With  $2\pi$  radians per sine wave cycle, the succeeding values in the table represent the sine values of 1 imes (6 $\pi$ /19), 2 imes (6 $\pi$ /19), 3 imes(6 $\pi$ /19),  $\ldots$  , up to 18 imes (6 $\pi$ /19). Consider the seventh and eighth values in the table, representing the sine values of 6  $\times$  (6 $\pi$ /19) and 7  $\times$  (6 $\pi$ /19) respectively. The respective calculatons of 16  $\times$  sin[6  $\times$  (6 $\pi$ /19)] and 16  $\times$  sin[7 imes (6 $\pi$ /19)] yield values of -5.20 and 9.83. Rounding to the nearest integer gives values of -5 and 10. When added to the base line value of 16, these values yield the results 11 and 26 for the seventh and eighth values in the 1336 Hz DTMF ROM table. Symmetry in the loop of 19 values in the DTME table dictates that the fourteenth and thirteenth values in the table are 21 and 6, representing values of 5 and -10 from the calculations.

The area under a half cycle of sine wave relative to the area of the surrounding rectangle is  $2/\pi$ , where  $\pi$  radians represent the sine wave half cycle. This surrounding rectangle has a length of  $\pi$  and a height of 1, with the height representing the maximum sine value. Consequently, the area of the surrounding rectangle is  $\pi$ . The integral of the area under the half sine wave from 0 to  $\pi$  is equal to 2. The ratio of  $2/\pi$  is equal to 63.66%, so that the total of the values for each half sine waves. The maximum values (relative to the base line) are 13 and 16 respectively for the low and high band DTMF frequencies.

For the previous 1336 Hz example, the total of the absolute values for the 19 sine values from the 1336 Hz ROM

table is equal to 196. The surrounding rectangle for the three cycles of sine wave is 19 by 16 for a total area of 304. The ratio of 196/304 is 64.47% compared with the  $2/\pi$  ratio of 63.66%. Thus the sine wave approximation gives an area abundance of 0.81% (equal to 64.47 - 63.66).

An application of the sine wave area criteria is shown in the generation of the DTMF 852 Hz frequency. The ten sine values calculated are 0, 7.64, 12.36, 12.36, 7.64, 0, -7.64, -12.36, -12.36, and -7.64. Rounding off to the nearest integer yields values of 0, 8, 12, 12, 8, 0, -8, -12, -12 and -8. The total of these values (absolute numbers) is 80, while the area of the surrounding rectangle is 130 (10 x 13). The ratio of 80/130 is 61.54% compared with the  $2/\pi$  ratio of 63.66%. Thus the sine wave approximation gives an area deficiency of 2.12% (equal to 63.66 - 61.54), which is overly deficient. Consequently, two of the ten sine values are augmented to yield sine values of 0, 8, 12, 13\*, 8, 0, -8, -12,  $-13^*$ , and -8. This gives an absolute total of 82 and a ratio of 82/130, which equals 63.08% and serves as a much better approximation to the  $2/\pi$  ratio of 63.66%.

The sine wave area criteria is also used to modify two values in the DTMF 941 Hz frequency. The nine sine values calculated are 0 8 36 12 80 11 26 4 45 -4 45 -11 26 12.80, and -8.36. Rounding off to the nearest integer yields values of 0, 8, 13, 11, 4, -4, -11, -13, and -8. The total of these values (absolute numbers) is 72, while the area of the surrounding rectangle is 117 (9 x 13). The ratio of 72/117 is 61.54% compared to the  $2/\pi$  ratio of 63.66%. Thus the sine wave approximation gives an area deficiency of 2.12% (equal to 63.66 - 61.54), which is overly deficient. Rounding up the two values of 4.45 and -4.45 to 5 and -5, rather than down to 4 and -4, yields values of 0, 8, 13, 11, 5, -5, -11, -13 and -8. This gives an absolute total of 74 and a ratio of 74/117, which equals 63.25% and serves as a much better approximation to the  $2/\pi$  ratio of 63.66%. With these modified values for the 852 and 941 DTMF frequencies, the area criteria ratio of  $2/\pi = 63.66\%$  for the sine wave compared to the surrounding rectangle has the following values:

DTMF Freg.	Sum of Values	Rectangle Area	Percentage	Diff.
697 Hz	406	$49 \times 13 = 637$	63.74%	+0.08%
770 Hz	92	11 x 13 = 143	64.34%	+0.68%
852 Hz	82	$10 \times 13 = 130$	63.08%	-0.58%
941 Hz	74	9 x 13 = 117	63.25%	-0.41%
1209 Hz	72	7 x 16 = 112	64.29%	+0.63%
1336 Hz	196	19 x 16 = 304	64.47%	+0.81%
1477 Hz	232	$23 \times 16 = 368$	63.04%	-0.62%
1633 Hz	216	21 x 16 = 336	64.29%	+0.63%

The LUP42 program loop is interrupted by the COP820C/ 840C timer after 100 ms of DTMF output. As stated previously, the Stack Pointer (SP) must be adjusted (incremented by 2) following the timer interrupt before returning from the DTMFLP subroutine.

This second DTMF generation subroutine for the COP820C/840C uses 301 bytes of code, consisting of 88 bytes of program code and 213 bytes of ROM table. The following is a summary of the DTMFLP subroutine code allocation.

DTMFLP Code	# of
Allocation	Bytes
1. Subroutine Header Code	42
2. Interrupt Code	16
3. LUP42 Code	30
4. Telephone Key Table	64
5. Sine Value Tables	149

301

Total

A program test routine to sequentially call the DTMFLP subroutine for each of the 16 DTMF keyboard input digits is supplied with the listing for the DTMF35 program. This test routine uses a 16 byte ROM table to supply the low true encoded column/row keyboard input to the accumulator. An input from the IO pin of the I port is u DTMF generation subroutine is to be subroutine is selected with I0 = 1.

A TYPICAL RESISTOR LADDER NETV FIGURF 4

# SUMMARY

In summary, the DTMF35 program assumes a COP820C/ 840C clocked with a 3.58 MHz crystal in divide by 10 mode. The DTMF35 program contains three subroutines, KBRDEC, DTMFGP, and DTMFLP. The KBRDEC subroutine is a low true DTMF keyboard decoder, while the DTMFGP and DTMFLP subroutines represent the alternative methods of DTMF generation.

The KBRDEC subroutine provides a low true decoding of the DTMF keyboard input and assumes that the keyboard input has been encoded in a low true column/row format, with the columns of the keyboard being sequentially strobed.

The DTMFGP subroutine produces two PWM (Pulse Width Modulation) outputs (representing the selected high and low band DTMF frequencies) for combination with an external op amp network (LM324 or equivalent).

The DTMFLP subroutine produces six bits of combined high band and low band DTMF frequency output for combination in an external resistor ladder network. This output represents a combined sine wave simulation of the two selected DTMF frequencies by combining values from two selected ROM tables, and updating these values every 118  $\mu$ s.

The three DTMF35 subroutines contain the following number of bytes of program and ROM table mem

the I port is used to select which	ber of bytes of program and ROM table memory:				
utine is to be used. The DTMFLF th $IO = 1$ .		<pre># of Bytes of Program</pre>	# of Bytes of ROM Table	Total # of Bytes	
ADDER NETWORK IS SHOWN IN	KBRDEC	19	16	35	
	DTMFGP	78	16	94	
	DTMFLP	88	213	301	
DTMF6 DTMF5 DTMF4 6 SINE WAVE OUTPUTS DTMF3 DTMF2 DTMF1	2R $R$ $=2R$ $R$ $R2R$ $R$ $R2R$ $R$ $R2R$ $R$ $R2R$ $R$ $R$		OUT		
FIGURE 4. Typical	Resistor Ladder Net	work	TL/DD/10740	)-24	

1	
2 3 4	DTMF GENERATION WITH A 3.58 MHZ VERNE H. WILSON CRYSTAL FOR COP820C/840C 10/28/89
* 5 6	DTMF - DUAL TONE MULTIPLE FREQUENCY
7	PROGRAM NAME: DTMF35.MAC
9	; .TITLE DTMF35
10	.CHIP 840;
12 13 14 15 16 17	; ; THIS DTMF PROGRAM IS BASED ON A COP820C/840C RUNNING ; WITH A CKI CLOCK OF 3.579545 MHZ (TV COLOR CRYSTAL ; FREQUENCY) IN DIVIDE BY 10 MODE, FOR AN INSTRUCTION ; CYCLE TIME OF 2.7936511 MICROSECONDS.
18 19 20 21 22	, THIS PROGRAM CONTAINS THREE SUBROUTINES, ONE FOR A ; LOW TRUE ROW/COLUMN DTMF KEYBOARD DECODING (KBRDEC), ; AND THE OTHER TWO (DTMFGP, DTMFLP) FOR ALTERNATE ; METHODS OF DTMF GENERATION.
23 24 25 26 27	; KEYBOARD INPUT DATA IS IN ACCUMULATOR WITH A ; LOW TRUE FORMAT AS FOLLOWS: ; BITS 7 TO 4 : LOW TRUE COLUMN VALUE (E,D,B,7) ; BITS 3 TO 0 : LOW TRUE ROW VALUE (E,D,B,7)
28 29 30	; ASSUMPTION MADE THAT COLUMN STROBES (LOW TRUE) ARE ; OUTPUT, WHILE ROW VALUES (LOW TRUE) ARE INPUT.
31 32 33 34 35 36	; THE FIRST METHOD OF DTMF GENERATION CONSISTS OF ; GENERATING TWO PWM OUTPUTS ON THE G PORT G2 AND G3 ; OUTPUT PINS. THESE TWO OUTPUTS NEED TO BE MIXED ; EXTERNALLY WITH AN APPROPIATE LM324 OP AMP FEEDBACK ; CIRCUIT TO GENERATE THE DTMF.
37 38 39 40 41 42 43	; THE SECOND METHOD OF DTMF GENERATION USES ROM LOOKUP ; TABLES TO SIMULATE THE TWO DTMF SINE WAVES AND ; COMBINES THEM ARITHMETICALLY. THE RESULT IS OUTPUT ON ; THE LOWER SIX BITS OF THE L PORT (LO - L5). THESE SIX ; OUTPUTS ARE COMBINED EXTERNALLY WITH A LADDER NETWORK ; TO GENERATE THE DTMF.
44 45 46 47 48 49	THE SECOND DTMF GENERATION METHOD USES APPROXIMATELY THREE TIMES AS MUCH ROM CODE (INCLUDING PROGRAM CODE AND ROM TABLES) AS THE FIRST METHOD, BUT HAS THE ADVANTAGE OF ELIMINATING THE COST OF THE EXTERNAL ACTIVE COMPONENT (LM324 OR EQUIVALENT).
49 50 51	; BOTH OF THE DTMF SUBROUTINES GENERATE THEIR OUTPUTS ; FOR A PERIOD OF 100 MILLISECONDS.
	TL/DD/10740-1

[			
52	;		
53	; DECLAR	RATIONS:	
54	;		
55 0000		KDATA = O	; *** KEYBOARD DATA ***
56 00D0		PORTLD = 0D0	; PORTL DATA REG
57 00D1		PORTLC = 0D1	; PORTL CONFIG REG
58 00D4		PORTGD = 0D4	; PORTG DATA REG
59 00D5		PORTGC = OD5	; PORTG CONFIG REG
60 00D7		PORTI = 0D7	; PORTI INPUT PINS
61 00DC		PORTD = ODC	; PORTD REG
62 00EA		TMRLO = OEA	; TIMER LOW COUNTER
63 00EB		TMRHI = OEB	: TIMER HIGH COUNTER
64 00EC		TAULO = OEC	; TMR AUTORELOAD REG LO
65 00ED		TAUHI = OED	; TMR AUTORELOAD REG HI
66 00EE		CNTRL = OEE	; CONTROL REG
67 00EF		PSW = OEF	; PROC STATUS WORD
68 00F0		RO = OFO	; LB FREQ LOOP COUNTER
		R0 = 0F0 R1 = 0F1	; LB FREQ LOOP COUNT
69 00F1			
70 00F2		R2 = 0F2	; LB FREQ Q COUNT
71 00F3		R3 = 0F3	; LB FREQ R COUNT
72	;		
73 0000 DD2F	START:	LD SP,#02F	; INITIALIZE STACK PTR
74	;		
75	;		KEYBOARD HEX DIGIT MATRIX
76	;		1 2 3 A
77 0002 DEDC		LD B, #PORTI	
78 0004 9E00		LD [B],#0	; 789C
79 0006 A0	LOOP:	RC	; * 0 # D
80 0007 AE		LD A,[B]	; DTMF TEST LOOP
81 0008 9405		ADD A,#5	; SEQUENCE IS 1,5,9,D,4,
82 000A A6		X A,[B]	; 8,#,A,7,0,3,B,*,2,6,C
83 000B 6C		RBIT 4,[B]	; HEX MATRIX TO LOOKUP
84 000C 9420		ADD A,#020	; TABLE FOR LOW TRUE
85 000E A4		LAID	; COLUMN/ROW INPUT TO
86 000F 3210		JSR KBRDEC	KBRDEC SUBROUTINE
87 0011 A1		SC	; SET C IF NOT SINGLE KEY
88 0012 DED7		LD B,#PORTI	
89 0014 70		IFBIT 0,[B]	; DETERMINE WHICH
90 0015 03		JP BYPA	; DTMF SUBROUTIINE
91 0016 3040			
92 0018 02		JSR DTMFGP	; TWO PWM OUTPUTS ON
	DVDA	JP BYPB	; G PORT PINS G2,G3
93 0019 308E	BYPA:	JSR DTMFLP	; SIX LADDER OUTPUTS ON
94			; L PORT PINS LO - L5
95 001B DEDC	BYPB:	LD B,#PORTI	
96 001D E8		JP LOOP	; CALL OF SUBROUTINE
97	;		
98	;		
99	;		
			TL/DD/10740-2
			12/20/10/40-2
1			

. FORM
; KEYBOARD DIGIT MATRIX TABLE
; . = 020
; ; 1 5 9 D 4 8 # A .BYTE OEE,ODD,OBB,077,0ED,ODB,0B7,07E
; 703B*26C
, .вуте о́ев, о́р7, о́ве, о́7р, о́е7, о́ре, о́вр, о́7в
;;;;;
; ; ; FIRST DTMF SUBROUTINE (DTMFGP) PRODUCES TWO PWM ; (PULSE WIDTH MODULATION) OUTPUTS ON PINS G3, G2 ;
; G PORT IS USED FOR THE TWO OUTPUTS ; - HIGH BAND (HB) FREQUENCY OUTPUT ON G3 ; - LOW BAND (LB) FREQUENCY OUTPUT ON G2
; TIMER COUNTS OUT ; - HB FREQUENCIES
; ; PROGRAM COUNTS OUT ; - LB FREQUENCIES ; - 100 MSEC DIVIDED BY LB HALF PERIOD QUOTIENT ; - 100 MSEC DIVIDED BY LB HALF PERIOD REMAINDER
; NOTE THAT ALL COUNTS MUST BE NORMALIZED TO THE ; 2.7936511 MICROSECOND INSTRUCTION CYCLE TC
; ; 100 msec represents 35796 Tc's ;
; TL/DD/1074

137	;
138 139 140 141	; ; HALF PERIODS FOR THE 8 DTMF FREQUENCIES (697,770,852, 941,1209,1336,1477, AND 1633 KHZ) ARE 257,232, 210,190,148,134,121, AND 110 Tc's RESPECTIVELY
142 143 144	; ; THE 100 MSEC DIVIDED BY HALF PERIOD QUOTIENTS ARE ; 139,154,170,188,241,267,295, AND 325 RESPECTIVELY
145 146 147 148 149	; ; THE 100 MSEC DIVIDED BY HALF PERIOD REMAINDERS ARE ; 72,67,95,75,127,17,100, AND 45 RESPECTIVELY ;
150 151 152 153 154 155	; ; BINARY FORMAT FOR THE HEX DIGIT KEY VALUE FROM THE ; KBRDEC SUBROUTINE IS OOOORRCC, ; WHERE - RR IS ROW SELECT (LB FREQUENCIES) - CC IS COLUMN SELECT (HB FREQUENCIES)
156 157 158	; ; FREQUENCY VECTORS (HB & LB) FOR FREQ PARAMETER TABLE ; MADE FROM KEY VALUE
159 160 161	; ; HB FREQ VECTORS (4) END WITH OO FOR TIMER COUNTS, ; WHERE VECTOR FORMAT IS OO11CCOO
162 163 164 165 166 167 168 169	; LB FREQUENCY VECTORS (12) END WITH: 11 FOR HALF PERIOD LOOP COUNTS, WHERE VECTOR FORMAT IS 0011RR11 10 FOR 100 MSEC DIVIDED BY HALF PERIOD QUOTIENTS, WHERE VECTOR FORMAT IS 0011RR10 11 FOR 100 MSEC DIVIDED BY HALF PERIOD REMAINDERS, WHERE VECTOR FORMAT IS 0011RR01
170 171 172	; ; FREQ PARAMETER TABLE AT HEX 003* (REQUIRED LOCATION) ;
173 174 175 176 177	KEY VALUE 0000RRCC
78 79 80 81 82	; TIMER T CCOO ; R1 F RR11 ; R2 Q RR10 ; R3 R RR01
83 84	
	TL/DD/10740

185		.FOF	RM	
186	;			
187	;		NCEC DADAMEMED	
188 189		AND 100	MSEC PARAMETER	IADLE
190 0030 93	;	. BYTE	147	; T
191 0031 0A		.BITE	10	; R
192 0032 8C		.BYTE	140	; Q
193 0033 26		.BYTE	38	; F
194 0034 85		.BYTE	133	; Т
195 0035 09		.BYTE	9	; R
196 0036 9B		.BYTE	155	; Q
197 0037 21		.BYTE	33	; F
198 0038 78		.BYTE	120	; T
199 0039 OE		.BYTE	14	; R
200 003A AB		.BYTE .BYTE	171 31	; Q ; F
201 003B 1F 202 003C 6D		.BITE	109	; T
202 003C 8D 203 003D 0A		.BITE	109	; R
204 003E BD		.BYTE	189	; Q
205 003F 1A		.BYTE	26	; F
206	;			
207	;			
208	;			
209 0040 DED5		LD	•	CONFIGURE G PORT
210 0042 9B3F		LD	[B-],#03F ;	
211 0044 6B		RBIT	- / L - J	OPTIONAL HB RESET
212 0045 6A		RBIT	-/	OPTIONAL LB RESET
213 0046 5F 214 0047 A6		LD X	B,#KDATA A,[B] ;	STORE KEY VALUE
214 0047 AB 215 0048 AE		LD		KEY VALUE TO ACC
216 0049 9733		OR		CREATE LB FREQ VECTOR
217 004B DEF1		LD	B,#R1 ;	
218 004D A6		X	A,[B]	
219 004E AE		LD	A,[B] ;	THREE PARAMETERS
220 004F A4		LAID	;	
221 0050 A2		х	A,[B+] ;	
222 0051 8B		DEC	A ;	TO R1,R2,R3
223 0052 44		IFBNE	#4	
224 0053 F9		JP	LUP B #KDAMA	
225 0054 5F 226 0055 AE		LD LD	B,#KDATA A,[B] ;	KEY VALUE TO ACC
227 0056 65		SWAP		CREATE HB FREQ VECTOR
228 0057 A0		RC		FROM KEY VALUE
229 0058 B0		RRC	A	
230 0059 B0		RRC	А	
231 005A 9730		OR	A,#030	
232 005C A4		LAID		HB FREQ TABLE
233 005D DEEA		LD		(1 PARAMETER)
234 005F 9A0F		LD		INSTRUCTION CYCLE
235 0061 9A00		LD	[B+],#0 ;	TIME UNTIL TOGGLE
				TL/DD/10740-5

236 0063 A2 237 0064 9A00 238 0066 9EB0 239 0068 DED4 240 006A DCF1 241 006C BB 242 006D 72 243 006E 03 244 006F B2 245 0070 7A 246 0071 03 247 0072 B8 248 0073 6A 249 0074 B2 250 0075 C2 251 0076 01 252 0077 0E 253 0078 C0 254 0079 FE 255 256 007A BE 257 007B 921F 258 007D EE	LUP1: BYP1: BYP2: LUP2: ;	X LD LD LD LD IFBIT JP X SBIT JP NOP RBIT X DRSZ JP DRSZ JP DRSZ JP LD IFEQ JP	A, [B+] [B+],#0 [B],#0B0 B,#PORTGD X,#R1 A, [X-] 2, [B] BYP1 A, [X+] 2, [B] BYP2 2, [B] A, [X+] R2 LUP2 FINI R0 LUP2 A, [X] A, #31 LUP1	<pre>; HB FREQ PARAMETER TO ; AUTORELOAD REGISTER ; START TIMER PWM ; TEST LB OUTPUT ; SET LB OUTPUT ; RESET LB OUTPUT ; DECR. QUOT. COUNT ; Q COUNT FINISHED ; DECR. F COUNT ; LB (HALF PERIOD) ; ************************************</pre>
255 256 007A BE 257 007B 921F	; FINI: ; ;	IFEQ	A,#31	; BALANCE ***
	,			TL/DD/10740-6

275		.FORM
276		
277		, SECOND DTMF SUBROUTINE (DTMFLP) PRODUCES SIX
278		COMBINED LOW BAND AND HIGH BAND FREQUENCY
279		SINE WAVE OUTPUTS ON PINS LO - L5
280		
281		, ; SIX L PORT OUTPUTS (LO - L5) FEED INTO AN EXTERNAL
282		; RESISTOR LADDER NETWORK TO CREATE THE DIMF OUTPUT.
283		, ADDIDION HADDAN ADIWONN IS CALAID THE DIMI COTTOL
284		, FOUR VALUES FROM A KEYBOARD ROM TABLE ARE LOADED
285		; INTO LFTBSZ (LOW FREQ TABLE SIZE), LFTADR (LOW
286		; FREQ TABLE ADDRESS), HFTBSZ (HIGH FREQ TABLE SIZE),
287		; AND HFTADR (HIGH FREQ TABLE ADDRESS).
288		, AND AFTADA (ATOM FALL TADAL ADDALSS).
289		, LUP42 USES THE LFPTR (LOW FREQ POINTER) AND HFPTR
290		; (HIGH FREQ POINTER) TO ACCESS THE SINE DATA TABLES
291		: FOR THE SELECTED FREQUENCIES ONCE PER LOOP. THESE
292		; POINTERS ARE BOTH INCREMENTED ONCE PER LUP42.
293		; FOINIERS ARE DOIN INCREMENTED ONCE FER HOFFE.
293		LUP42 PROGRAM LOOP UPDATES THE OUTPUT VALUE EVERY
295		; 117 1/3 USEC BY SELECTING AND THEN COMBINING NEW
296		; VALUES FROM THE SELECTED LOW BAND AND HIGH BAND
297		: FREQUENCY ROM TABLES WHICH SIMULATE THE SINE WAVES
298		; FOR THE TWO FREQUENCIES.
298		; FOR THE INO FREQUENCIES.
		; MULTIPLES OF THE MAGIC NUMBER OF APPROXIMATELY
300		; 118 USEC ARE CLOSE APPROXIMATIONS TO ALL EIGHT OF
301		
302		; THE DTMF FREQUENCIES.
303		; COP820C/840C TIMER USED TO INTERRUPT THE DTMF LUP42
304		
305		; PROGRAM LOOP AFTER 100 MSEC TO FINISH THE DTMF
306		; OUTPUT AND RETURN FROM THE DTMFLP SUBROUTINE. NOTE
307		; THAT THE STACK POINTER (SP) MUST BE ADJUSTED AFTER
308		; THE INTERRUPT BEFORE RETURNING FROM THE SUBROUTINE.
309		
310		;
311		;
312		:
313		
314		; DECLARATIONS:
315		
316	0005	LFPTR = 05 ; LOW FREQ POINTER
317	0006	TEMP = 06 ; TEMPORARY
318	0007	HFPTR = 07 ; HIGH FREQ POINTER
319	0008	LFTBSZ = 08 ; LO FREQ TABLE SIZE
320	0009	LFTADR = 09 ; LO FREQ TABLE ADDR
321	000A	HFTBSZ = OA ; HI FREQ TABLE SIZE
322	000B	HFTADR = OB ; HI FREQ TABLE ADDR
323		;
324	0004	TRUN = 04
325		;
		TI /DD/10740-7
		TL/DD/10/40-7
1		

328 329 330 331 332 333 334 335 336 337 338	008E BCD1FF 0091 BCD01D 0094 BC0500 0097 58 0098 9A00 009A A0 009B 65 009C B0 009D B0 009E 94B8 00A0 A6 00A1 AE 00A2 A4	; DTMFLP: FRLUP:	LD LD LD LD RC SWAP RRC RRC ADD X LD LAID	PORTLC, #OFF PORTLD, #29 LFPTR,#0 B,#HFPTR [B+],#0 A A A,#0B8 A,[B] A,[B]	; INITIALIZE PORT L ; FOR NO TONE OUT ; INITIALIZE OFFSET ; POINTERS FOR ; DTMF SINE WAVE ; TABLE LOOKUP ; QUADRUPLE KEY ; VALUE AND ADD ; OFFSET FOR KEY ; TABLE LOOKUP ; LOAD FOUR VALUES ; FROM ROM KEY ; TABLE INTO LOW
341 342 343 344 345 346 347 348	00A3 A2 00A4 8A 00A5 4C 00A6 F9 00A7 DEEA 00A9 9A00 00AB 9A8C 00AD 9A00 00AF 9A8C 00B1 9A80		X INC JFBNE LD LD LD LD LD LD	A,[B+] A #OC FRLUP B,#TMRLO [B+],#O [B+],#140 [B+],#140 [B+],#140 [B+],#080	; FREQ LFTBSZ, ; LFTADR, AND HI ; FREQ HFTBSZ, ; HFTADR ; INITIALIZE TIMER ; WITH A tC COUNT ; EQUIVALENT TO ; 100 MSEC PLUS ; A LUP42 TIME ; TIMER PWM, NO OUT
350 351	00B3 9B11 00B5 7C 00B6 210F	; ; ; TELEPH(	LD SBIT JMP DNE KEY TAB	[B-],#011 TRUN,[B] LUP42	; ENABLE TMR INTRPT ; START TIMER
359 360 361 362 363 364 365 366	00B8 31	: TAB] ; ; ; ; ; KEY 1	PARAMETER PARAMETER	2: BASE ADDR. 3: # OF HIGH	REQ TABLE VALUES OF LOW FREQ VALUES FREQ TABLE VALUES OF HIGH FREQ VALUES
367 368		; ; key 2			
369 370	00BC 31 00BD 2D 00BE 13 00BF 83	;	. BYTE	<b>49,02D,19,08</b> 3	
					TL/DD/10740-8

371       ; KEY 3       .BTE       49,02D,23,096         373       ; KEY A       .BTE       49,02D,21,0AD         374       ; KEY A       .BTE       49,02D,21,0AD         375       ; OCC 31       .BTE       11,05E,7,07C         00C6 13       .BTE       .BTE       11,05E,7,07C         00C6 07       .BTE       .BTE       11,05E,19,083         00CC 08       .BTE       .BTE       11,05E,23,096         00CC 13       .BTE       .BTE       11,05E,23,096         00CE 13       .BTE       .BTE       11,05E,21,0AD         00D1 08       .EEY 6       .BTE       .BTE       10,069,7,07C         00D5 15       .BTE       .BTE       10,069,7,07C       .BTE       .BTE       10,069,19,083         00D5 13       .BTE       .BT							
373       :       :       KEY A         375       :       KEY A         375       :       KEY A         9006       15         9007       AD         376       :         377       :         378       00000         90000       5E         00000       00000         00000       5E         000000       5E		00C1 2D 00C2 17	;	КЕҮ З	. BYTE	49,02D,23,096	
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	374	00C4 31 00C5 2D 00C6 15		KEY A	.BYTE	49,02D,21,0AD	
379       :       KEY 5         380       :       KEY 5         381       00CC 0B       .BYTE       11,05E,19,083         00CF 83       .BYTE       11,05E,23,096         382       :       KEY 6         383       :       KEY 6         384       00D 0B       .BYTE       11,05E,23,096         00D1 5E       .BYTE       11,05E,21,0AD         00D5 5E       .BYTE       11,05E,21,0AD         00D5 5E       .BYTE       11,05E,21,0AD         00D5 15       .BYTE       10,069,7,07C         00D4 0B       .BYTE       10,069,7,07C         00D5 7C       .BYTE       10,069,19,083         00D6 13       .BYTE       10,069,19,083         00D5 7C       .BYTE       10,069,19,083         00D5 7C       .BYTE       10,069,19,083         00D5 63       .BYTE       10,069,19,083         00D5 7C       .BYTE       10,069,19,083         00D5 7B3       .BYTE       10,069,23,096	377	00C8 0B 00C9 5E 00CA 07	;;	KEY 4	. BYTE	11,05E,7,07C	
383       ; KEY 6         384       00D0 0B       .BYTE       11,05E,23,096         00D1 5E       .00D2 17         00D3 96       .EYTE       11,05E,23,096         385       ; KEY B       .BYTE         386       ; KEY B       .BYTE         00D5 5E       .BYTE       11,05E,21,0AD         00D7 AD       .BYTE       10,069,7,07C         388       ; KEY 7       .BYTE         390       00B8 0A       .BYTE       10,069,7,07C         00D9 69       .BYTE       10,069,19,083         00DD 69       .BYTE       10,069,19,083         391       ; KEY 8       .BYTE         393       00DC 0A       .BYTE       10,069,23,096         394       ; KEY 9       .BYTE       10,069,23,096	380 381	00CD 5E 00CE 13	;	KEY 5	.BYTE	11,05E,19,083	
386       ; KEY B         387       004 0B       .BYTE       11,05E,21,0AD         00D5       5E       .BYTE       11,05E,21,0AD         00D6       15       .BYTE       10,069,7,07C         00D9       69       .BYTE       10,069,7,07C         00D8       0A       .BYTE       10,069,19,083         391       ;       KEY 8         392       ;       KEY 8         393       00DC 0A       .BYTE       10,069,19,083         00DF       83       .BYTE       10,069,19,083         394       ;       .BYTE       10,069,23,096	383 384	00D1 5E 00D2 17	;	KEY 6	. BYTE	11,05E,23,096	
389       ; KEY 7         390       00D8       0A       .BYTE       10,069,7,07C         00D9       69       .00D8       7C         391       ;       .BYTE       10,069,7,07C         391       ;       .BYTE       10,069,7,07C         391       ;       .BYTE       10,069,19,083         392       ;       KEY 8         393       00DC 0A       .BYTE       10,069,19,083         00DF 13       .00DF 83       .BYTE       10,069,23,096         394       ;       .BYTE       10,069,23,096	386 387	00D5 5E 00D6 15	;	КЕҮ В	. BYTE	11,05E,21,0AD	
392     ; KEY 8       393     00DC 0A     .BYTE       00Db     69       00DF     83       394     ;       395     ; KEY 9       396     00E0 0A       00E1 69     .BYTE	389 390	00D9 69 00DA 07	;	KEY 7	. BYTE	10,069,7,07C	
395 ; KEY 9 396 00E0 0A .BYTE 10,069,23,096 00E1 69	392 393	00DD 69 00DE 13	;;	KEY 8	. BYTE	10,069,19,083	
	395		;;	KEY 9	. BYTE	10,069,23,096	TL/DD/10740-9

	00E2	17					
	00E2						
397	0010		;				
398			, KEY	С			
	00E4	0 A	, _		.BYTE	10,069,21,0AD	
	00E5	69					
	00E6						
400	00E7	AD					
400 401			; ; key	*			
	00E8	09	, 151		BYTE	9,073,7,083	
	00E9						
	OOEA	07					
	OOEB	83					
403			; ; key	n			
404	00EC	09	; KEI		BYTE	9,073,19,07C	
100	OOED					3,010,13,010	
	OOEE						
	OOEF						
406			;				
407	0070	00	; KEY		DVMD	0 070 00 001	
408	00F0 00F1				. BYTE	9,073,23,096	
	00F1						
	00F3						
409		-	;				
410			; KEY	D			
411	00F4				BYTE	9,073,21,0AD	
	00F5						
	00F6 00F7						
412	001/	<b>Α</b> <i>μ</i>	;				
413			;				
414			;				
415			;				
416		OOFF			=00FF		
417	0055	BCD01D	; ТМФРDФ		'n	DOD#10 #20	; BASE LINE VALUE
	0102		INTRPT		D D	PORTLD,#29 B,#PSW	; 100 MSEC INTERRUPT
	0104				.D	[B-],#0	; FROM TIMER
	0106				D	[B],#O	; CLR PSW AND CNTRL
	0108				D	B,#SP	; RESTORE STACK
	010A				D	A,[B]	; POINTER (SP)
	010B 010C				INC INC	A A	; TO ITS VALUE ; BEFORE THE
	010C				( (	A A,[B]	; BEFORE THE ; INTERRUPT
	010E				Ret	,[~]	; RETURN FROM
<b>128</b>	-			-			; SUBROUTINE
129			;				
130			;				
							TL/DD/10740-

431	.FORM		
432 433	; LUP42 CONSISTS	OF 42 COP840C 1	NSTRUCTION CYCLE TIMES
434			79545 = 117 1/3  uSEC
435	; 10142 111110 10		
436			
437 010F 5A	LUP42: LD	B,#LFPTR	
438 0110 AE	LD	A,[B]	; INCREMENT LOW FREQ
439 0111 8A	INC	Α	; OFFSET POINTER
440 0112 57	LD	B,#LFTBSZ	; TEST IF LFPTR
441 0113 82	IFEQ	A,[B]	; BEYOND LIMIT
442 0114 64	CLR	A	; REINITIALIZE LFPTR
443 0115 5A	LD	B,#LFPTR	; FOR NEXT TIME
444 0116 A6	X	A,[B]	
445 0117 56	LD	B, #LFTADR	; ADD PTR TO LO FREQ
446 0118 84	ADD	A,[B]	; TABLE ADDRESS
447 0119 A4	LAID	5 485115	; LOW FREQ COMPONENT
448 011A 59	LD	B,#TEMP	; RESULT TO TEMP
449 011B A2	X	A,[B+]	; INCREMENT HI FREQ
450 011C AE		A,[B]	
451 011D 8A 452 011E 55	INC LD	A B,#HFTBSZ	; OFFSET POINTER ; TEST IF HFPTR
452 011E 55 453 011F 82	LD IFEQ	B,#HFTB52 A,[B]	; IESI IF HFFIR ; BEYOND LIMIT
453 0117 82	CLR	A,[D] A	; REINITIALIZE HEPTR
455 0121 58	LD	B,#HFPTR	; FOR NEXT TIME
456 0122 A6	X	A,[B]	, ton abit time
457 0123 54	LD	B,#HFTADR	; ADD PTR TO HI FREQ
458 0124 84	ADD	A,[B]	TABLE ADDRESS
459 0125 A4	LAID		; HI FREQ COMPONENT
460 0126 59	LD	B,#TEMP	; ADD LOW FREQ VALUE
461 0127 84	ADD	A,[B]	; TO HI FREQ VALUE
462 0128 9CD0	х	A, PORTLD	; RESULT TO PORT L
463 012A A4	LAID		; EQUIVALENT OF
464 012B A4	LAID		; SIX NOP'S
465 012C E2	JP	LUP42	; TIMING LOOP OF
466			; 117 1/3 uSEC
467	;		
468	;		
469	;		
470	;		
			TL/DD/10740-

473	FORM
471	. FORM
472	;
473	; THE FREQUENCY APPROXIMATION IS EQUAL TO THE NUMBER OF
474	; CYCLES OF SINE WAVE DIVIDED BY THE TIME IN THE TOTAL
475	; NUMBER OF LUP42 CYCLES BEFORE THE REPETITION OF THE
476	; ROM TABLE. AS AN EXAMPLE, CONSIDER THE THREE CYCLES
477	OF SINE WAVE AND 19 VALUES IN THE ASSOCIATED 1336 HZ
478	ROM TABLE. THE 19 CYCLES OF LUP42 TIMES THE LUP42
479	; TIME OF 117 1/3 USEC IS DIVIDED INTO THE THREE CYCLES
	; OF SINE WAVE TO YIELD A VALUE OF 1345.69 HZ AS THE
480	
481	; 1336 HZ APPROXIMATION.
482	;
483	THE VALUES IN THE ROM TABLES FOR THE DTMF SINE WAVES
484	; SHOULD WRAP AROUND END TO END IN EITHER DIRECTION TO
485	; FORM A SYMETRICAL LOOP. THE FIRST VALUE IN THE ROM
486	; TABLE REPRESENTS THE BASE LINE FOR THAT FREQUENCY.
487	
488	, THE HIGH BAND DTMF FREQUENCIES HAVE A BASE LINE VALUE
400	; OF 16 AND A MAXIMUM VALUE OF 32. THE LOW BAND DIMF
490	; FREQUENCIES HAVE A BASE LINE VALUE OF 13 AND A
491	; MAXIMUM VALUE OF 26. THIS DIFFERENCE IN BASE LINE
492	; VALUES IS NECESSARY TO SATISFY THE REQUIREMENT OF THE
493	; HIGH BAND FREQUENCIES NEEDING A LEVEL 2 OB ABOVE THE
494	; LEVEL OF THE LOW BAND FREQUENCIES TO COMPENSATE FOR
495	: LOSSES IN TRANSMISSION. THE SUM OF THE TWO BASE LINE
496	; VALUES YIELDS A BASE LINE VALUE OF 29, WHILE THE SUM
497	; OF THE TWO MAXIMUM VALUES YIELDS A MAXIMUM VALUE OF
	; 58. THUS THE SIX BIT DTMF OUTPUT FROM THE L PORT TO
498	
499	; THE LADDER NETWORK RANGES FROM 0 TO 58, WITH A BASE
500	; LINE VALUE OF 29.
501	
502	; THE VALUES IN THE DTMF SINE WAVE TABLES ARE
503	: CALCULATED BY COMPUTING THE SINE VALUE AT THE
504	; APPROPIATE POINTS, SCALING THE SINE VALUE UP TO THE
505	; BASE LINE VALUE, AND THEN ADDING THE RESULT TO THE
506	; BASE LINE VALUE. THE FOLLOWING EXAMPLE WILL HELP TO
507	; CLARIFY THIS CALCULATION.
508	;
509	; CONSIDER THE THREE CYCLES OF SINE WAVE ACROSS 19
510	; DATA POINTS FOR THE 1336 HZ DTMF HIGH BAND FREQUENCY.
511	; THE FIRST VALUE IN THE TABLE IS THE BASE LINE VALUE
512	; OF 16. WITH 2 PI RADIANS PER SINE WAVE CYCLE,
513	; THE SUCCEEDING VALUES IN THE TABLE REPRESENT THE
514	; SINE VALUES OF 1 X (6 PI / 19), 2 X (6 PI / 19),
515	; 3 X (6 PI / 19), , UP TO 18 X (6 PI / 19).
516	; LET US NOW CONSIDER THE SEVENTH AND EIGHTH VALUES
517	; IN THE TABLE, REPRESENTING THE SINE VALUES OF
518	; 6 X (6 PI / 19) AND 7 X (6 PI / 19) RESPECTIVELY.
519	; THE CALCULATIONS OF 16 X SIN [6 X (6 PI / 19)] AND
520	; 16 X SIN [7 X (6 PI / 19)] YIELD VALUES OF - 5.20 AND
521	9.83 RESPECTIVELY. ROUNDED TO THE NEAREST INTEGER
	; THE CALCULATIONS OF 16 X SIN [6 X (6 PI / 19)] AND ; 16 X SIN [7 X (6 PI / 19)] YIELD VALUES OF - 5.20 AND ; 9.83 RESPECTIVELY. ROUNDED TO THE NEAREST INTEGER TL/DD/10740-12
	TL/DD/10740-12
	00
	20

523       ;         524       ;         525       ;         526       ;         527       ;         528       ;         529       ;         530       ;         531       ;         532       ;         533       ;         534       ;         535       ;         536       ;         537       ;         538       ;         539       ;         540       ;         541       ;	GIVES VALUES OF - 5 AND 10. WHEN ADDED TO THE BASE LINE VALUE OF 16, THESE VALUES YIELD THE RESULTS 11 AND 26 FOR THE SEVENTH AND EIGHTH VALUES IN THE 1336 HZ DTMF TABLE. SYMMETRY IN THE LOOP OF 19 VALUES IN THE DTMF TABLE DICTATES THAT THE FOURTEENTH AND THIRTEENTH VALUES IN THE TABLE ARE 21 AND 6, REPRESENTING VALUES OF 5 AND - 10 FROM THE CALCULATIONS. THE AREA UNDER A HALF CYCLE OF SINE WAVE RELATIVE TO THE AREA OF THE SURROUNDING RECTANGLE IS 2/PI, WHERE PI RADIANS REPRESENT THE SINE WAVE HALF CYCLE. THIS SURROUNDING RECTANGEE HAS A LENGTH OF PI AND A HEIGHT OF 1, WITH THE HEIGHT REPRESENTING THE MAXIMUM SINE VALUE. CONSEQUENTLY, THE AREA OF THIS SURROUNDING RECTANGLE IS PI. THE INTEGRAL OF THE AREA UNDER THE HALF SINE WAVE FROM 0 TO PI IS EQUAL TO 2. THE RATIO OF 2/PI IS EQUAL TO 63.66 %, SO THAT THE TOTAL OF THE VALUES FOR EACH HALF SINE WAVE SHOULD APPROXIMATE 63.66 % OF THE SUM OF THE MAX VALUES. THE MAXIMUM VALUES (RELATIVE TO THE BASE LINE) ARE 13 AND 16 RESPECTIVELY, FOR THE LOW AND HIGH BAND FREQUENCIES.
548 ; 549 ;	LF697: 4 CYCLES OF SINE WAVE SPREAD
550 ; 551 ;	ACROSS 49 TIMING LOOP (LUP42) CYCLES FREQ. = 4 / (49 X 117 1/3) = 695.73 HZ
552 ; 553 ; 554 ;	ERROR = $(697 - 695.73) / 697 = -0.18 \%$
554 555 012D 0D 012E 13 012F 18 0130 1A 0131 19 0132 14 0133 0E 0134 07 0135 02 0136 00 556 0137 01 0138 05 0139 0B 013A 12 013B 17 013C 1A 013D 19 013E 15	.BYTE 13,19,24,26,25,20,14,7,2,0 .BYTE 1,5,11,18,23,26,25,21,15,9
	TL/DD/10740-13

	013F 0F				
	0140 09		BYTE	3,0,1,4,10,16,22,25,26,23	
	0141 03 0142 00		.DIIE	5,0,1,4,10,10,22,25,20,25	
	0143 01				
	0144 04				
	0145 0A				
	0146 10				
	0147 16				
	0148 19 0149 1A				
	0144 17				
	014B 11		BVWD	17,11,5,1,0,3,8,15,21,25	
	014C 0B		.DIIL	17,11,5,1,0,5,0,15,21,25	
	014D 05				
	014E 01				
	014F 00				
	0150 03				
	0151 08				
	0152 OF				
	0153 15				
	0154 19				
	0155 1A		BYTE	26,24,19,12,6,1,0,2,7	
	0156 18			,,_,_,_,_,,,,,,,,,,,,,,,,,,,,,,,,,,	
	0157 13				
	0158 OC				
	0159 06				
	015A 01				
	015B 00				
	015C 02				
	015D 07				
560		;			
561		;			
562		; LF770:	1 CYCLE	OF SINE WAVE SPREAD	
563		;		ACROSS 11 TIMING LOOP (LUP42) C	YCLES
564		;			
565		;		1 / (11 X 117 1/3) = 774.79 HZ	
566		;		(774.79 - 770) / 770 = + 0.62	%
567	015B 05	;	B		
	015E 0D		.BITE	13,20,25,26,23,17,9,3,0,1	
	015F 14				
	0160 19				
	0161 1A				
	0162 17				
	0163 11				
	0164 09				
	0165 03				
	0166 00				
	0167 01		DVMP	6	
569 570	0168 06		.BYTE	6	
570		;			
					TL/DD/10740-14

571 572 573 574 575	; LF852: 1 CYCLE OF SINE WAVE SPREAD ACROSS 10 TIMING LOOP (LUP42) CYCLES ; FREQ. = 1 / (10 X 117 1/3) = 852.27 HZ
576 577 0169 0D 016A 15 016B 19 016C 1A 016D 15 016E 0D 016F 05 0170 01 0171 00	; ERROR = (852.27 - 852) / 852 = + 0.03 % ; .BYTE 13,21,25,26,21,13,5,1,0,5
0172 05 579 580 581 582 583 583 584 585	; ; LF941: 1 CYCLE OF SINE WAVE SPREAD ; ACROSS 9 TIMING LOOP (LUP42) CYCLES ; ; FREQ. = 1 / (9 X 117 1/3) = 946.97 HZ ; ERROR = (946.97 - 941) / 941 = + 0.63 %
586 587 0173 0D 0174 15 0175 1A 0176 18 0177 12 0178 08 0179 02 017A 00 017B 05	; .BYTE 13,21,26,24,18,8,2,0,5
588 589 590 591 592 593 593	; ; HF1209: 1 CYCLE OF SINE WAVE SPREAD ACROSS 7 TIMING LOOP (LUP42) CYCLES ; ; FREQ. = 1 / (7 X 117 1/3) = 1217.53 HZ
595 596 597 017C 10 017D 1D 017E 20 017F 17 0180 09 0181 00 0182 03	; ERROR = (1217.53 - 1209) / 1209 = + 0.71 % ; .BYTE 16,29,32,23,9,0,3
598	; TL/DD/10740-

599	;
600 601	; HF1336: 3 CYCLES OF SINE WAVE SPREAD : ACROSS 19 TIMING LOOP (LUP42) CYCLES
602	; ACROSS 19 TIMING LOOP (LOP42) CICLES
603	; FREQ. = 3 / (19 X 117 1/3) = 1345.69 HZ
604	; ERROR = (1345.69 - 1336) / 1336 = + 0.73 %
605 606 0183 10	; BYTE 16,29,31,19,4,0,11,26,32,24
0184 1D	.5115 10,23,01,13,1,0,11,20,02,21
0185 lF	
0186 13	
0187 04 0188 00	
0189 OB	
018A 1A	
018B 20	
018C 18 607 018D 08	.BYTE 8,0,6,21,32,28,13,1,3
018E 00	·biii 0,0,0,21,02,20,10,1,0
018F 06	
0190 15	
0191 20 0192 1C	
0192 IC 0193 0D	
0194 01	
0195 03 608	
609	
610	; HF1477: 4 CYCLES OF SINE WAVE SPREAD
611	; ACROSS 23 TIMING LOOP (LUP42) CYCLES
612 613	; ; FREQ. = 4 / (23 X 117 1/3) = 1482.21 HZ
614	; ERROR = $(1482.21 - 1477) / 1477 = + 0.35 \%$
615	;
616 0196 10 0197 1E	<b>.BYTE</b> 16,30,29,14,1,4,20,32,26,10
0197 1E 0198 1D	
0199 OE	
019A 01	
019B 04 019C 14	
019D 20	
019E 1A	
019F 0A 617 01A0 00	RYTTE 0 8 24 32 22 6 0 12 28 31
01A0 00 01A1 08	<b>.BYTE</b> 0,8,24,32,22,6,0,12,28,31
01A2 18	
01A3 20	
01A4 16 01A5 06	
01A5 00 01A6 00	
	TL/DD/10740-

	01A7					
	01A8 01A9					
618	OIAA			. BYTE	18,3,2	
	Olab					
	01AC	02				
619			;			
620 621			; 	A CYCLES	OF SINE WAVE SPREAD	
622			:	4 CICDER	ACROSS 21 TIMING LOOP (LUP42) CYCLI	ES
623			;			
624			;		4 / (21 X 117 1/3) = 1623.38 HZ	
625			;	ERROR =	(1633 - 1623.38) / 1633 = -0.59	*
626 627	OIAD	10	;	BYTE	16,31,27,9,0,11,29,30,14,0	
02.	OIAE				10,01,0,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	
	Olaf	1 B				
	01B0					
	0181					
	01B2 01B3					
	01B4					
	0185					
	01B6					
628	0187			.BYTE	7,25,32,18,2,3,21,32,23,5	
	01B8					
	01B9 01BA					
	OIBB					
	01BC					
	OlBD					
	OIBE					
	01BF 01C0					
629	0100			. BYTE	1	
630			;		-	
631			;			
632			;			
					TL/I	DD/10740-17

633	.FORM
	;
635	: DTMF KEYBOARD DECODE SUBROUTINE (KBRDEC)
636	, DIMI KLIDOMID DICCIL CODMONTAL (ADDIC)
637	, KEYBOARD INPUT DATA IS IN ACCUMULATOR WITH A
638	: LOW TRUE FORMAT AS FOLLOWS:
639	BITS 7 TO 4 : LOW TRUE COLUMN VALUE (E,D,B,7)
640	: BITS 3 TO 0 : LOW TRUE ROW VALUE $(E,D,B,7)$
641	, BIIS 5 TO 0 . HOW TRUE NOW VALUE (E, D, D, T)
642	, ASSUMPTION MADE THAT COLUMN STROBES (LOW TRUE) ARE
643	; OUTPUT, WHILE ROW VALUES (LOW TRUE) ARE INPUT.
644	
645	; LOW TRUE COLUMN/ROW INPUT DIGIT IN ACCUMULATOR IS
646	; TRANSFORMED INTO A DTMF HEX DIGIT KEY VALUE
647	;
648	; TABLE LOOKUP TRANSFORMATION CHECKS FOR MULTIPLE KEYS,
649	; NO KEY, OR NO COLUMN SELECT, AND THEN PRODUCES
650	; A DTMF HEX DIGIT KEY VALUE WITH A BINARY FORMAT
651	; OF 0000RRCC FOR A SINGLE KEY INPUT,
652	; WHERE - RR IS LOW BAND (LB) FREQUENCY SELECT
653	; - CC IS HIGH BAND (HB) FREQUENCY SELECT
654	;
655	; KBRDEC SUBROUTINE IS EXITED WITH A RETURN (RET)
656	COMMAND TO INDICATE MULTIPLE KEYS, NO KEY,
657	: OR NO COLUMN SELECT
658	;
659	; KBRDEC SUBROUTINE IS EXITED WITH A RETURN AND SKIP
660	(RETSK) COMMAND TO INDICATE A SINGLE KEY ENTRY
661	, (
662	;
663 0200	,=0200
664	
	, ; LOW TRUE TRANSLATION TABLE - ONLY E,D,B,7 ACCEPTABLE
-	; DOW INCE INANGENTION INDEE ONET E, D, D, F RECEITADED
667 0200 C0	, .BYTE 0C0,0C0,0C0,0C0,0C0,0C0,0C0,0C
0201 C0	
0202 C0	
0203 C0	
0204 C0	
0205 C0	
0206 C0	
0207 OC	
668 0208 C0	.BYTE 0C0,0C0,0C0,8,0C0,4,0,0C0
0209 C0	
020A C0	
020B 08	
020C C0	
020D 04	
020E 00	
020F C0	
	;
	TL/DD/10740-18

670	;			
671 0210 5F	KBRDEC:	LD	B,#KDATA	
672 0211 A6		Х		; STORE LOW TRUE
673 0212 AE		LD		; COLUMN/ROW VALUE
674 0213 95F0		AND		; EXTRACT LOW TRUE COLUMN
675 0215 65		SWAP	A	; & PUT IN LOWER NIBBLE
676 0216 A4		LAID		; OOOOCCOO FROM TABLE
677 0217 A0		RC		; SHIFT TABLE VALUE DOWN
678 0218 B0		RRC	Α	; TWO BITS TO PRODUCE
679 0219 B0		RRC	Α	; 000000CC
680 021A A6		Х	A,[B]	; STORE RESULT
681 021B 950F		AND		; EXTRACT LOW TRUE ROW
682 021D A4		LAID		; OOOORROO FROM TABLE
683 021E 84		ADD		; ADD TO PRODUCE OOOORRCC
684 021F 930F		IFGT		; RETURN IF MULTIPLE KEYS,
685 0221 8E		RET		; NO KEYS, OR NO COLUMN
686 0222 8D		RETSK		; RETURN AND SKIP
687				; IF SINGLE KEY
688	;			
689	;			
690	;			
691		. END		
				TL/DD/10740-19

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DTMF Generation with a 3.58 MHz Crystal	B BYPB FINI HFTBSZ LFPTR PORTD PORTLC R1 START TMRHI	0005 004D 00DC	CNTRL 00 FRLUP 00 INTRPT 00 LUP1 00 PORTGC 00 PORTLD 00 R2 00 TAUHI 00	DEE         1           DAO         1           DFF         1           DOFO         1           DOEC         1           DD5         1           DD5         1           DD7         1	BYP2 DTMFGP HFPTR KBRDEC LFTBSZ LUP2 PORTGD PSW R3 TAULO TRUN	0007 0210 0008 0078	BYPA DTMFLP HFTADR KDATA LOOP LUP42 PORTI RO SP TEMP X		-20
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