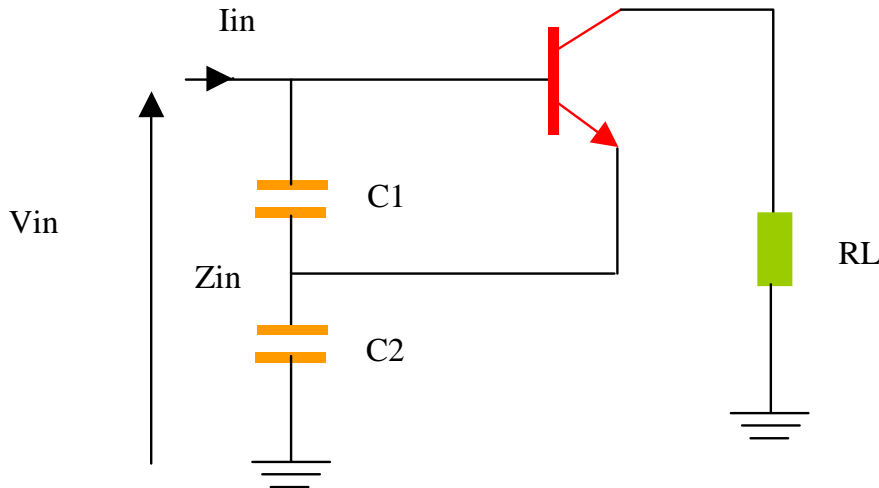


Derivation of Colpitts feedback capacitor values.



Simplified schematic diagram showing the reflection amplifier part of the Colpitts oscillator. The capacitor network of C1 and C2 would provide positive feedback from the emitter to base.

$$V_{in} = I_{in}(XC_1 + XC_2) - I_b(XC_1 - \beta XC_2) \quad - (1)$$

$$0 = -I_{in}(XC_1) + I_b(XC_1 + hie) \quad - (2)$$

rearrange to give I_b and sub into equation (1) ie $I_b = \frac{I_{in} \cdot XC_1}{(XC_1 + hie)}$

$$V_{in} = I_{in}(XC_1 + XC_2) - \frac{I_{in} \cdot XC_1}{(XC_1 + hie)}(XC_1 - \beta XC_2) \quad \text{and multiply out}$$

$$V_{in} = I_{in} \cdot XC_1 + I_{in} \cdot XC_2 - \frac{I_{in} \cdot XC_1}{(XC_1 + hie)} \cdot XC_1 + \frac{I_{in} \cdot XC_1}{(XC_1 + hie)} \cdot \beta XC_2 \quad \text{x both sides by } XC_1 + hie$$

$$V_{in}(XC_1 + hie) = (XC_1 + hie)I_{in} \cdot XC_1 + (XC_1 + hie)I_{in} \cdot XC_2 - (XC_1 + hie) \cdot \frac{I_{in} \cdot XC_1}{(XC_1 + hie)} \cdot XC_1 + (XC_1 + hie) \frac{I_{in} \cdot XC_1}{(XC_1 + hie)} \cdot \beta XC_2 \quad \text{and multiply out again.}$$



$$V_{in}(XC_1 + hie) = I_{in}XC_1^2 + hie.I_{in}.XC_1 + XC_1.I_{in}.XC_2 + hie.I_{in}.XC_2 - I_{in}XC_1^2 + XC_1.I_{in}.XC_2.\beta$$

$$V_{in}(XC_1 + hie) = hie.I_{in}.XC_1 + XC_1.I_{in}.XC_2 + hie.I_{in}.XC_2 + XC_1.I_{in}.XC_2.\beta \text{ Rearrange for } I_{in}$$

$$V_{in}(XC_1 + hie) = I_{in}[hie.(XC_1 + XC_2) + XC_1.XC_2.I_{in}(1 + \beta)]$$

$$\frac{V_{in}}{I_{in}} = Z_{in} = \frac{hie.(XC_1 + XC_2) + XC_1.XC_2(1 + \beta)}{XC_1 + hie}$$

If we assume that $XC_1 \ll hie$ then :-

$$\frac{V_{in}}{I_{in}} = Z_{in} = \frac{hie.(XC_1 + XC_2) + XC_1.XC_2(1 + \beta)}{hie} \Rightarrow \frac{hie.(XC_1 + XC_2)}{hie} + \frac{XC_1.XC_2(1 + \beta)}{hie}$$

$$\frac{V_{in}}{I_{in}} = Z_{in} = XC_1 + XC_2 + \frac{XC_1.XC_2(1 + \beta)}{hie} \text{ Let } gm = \frac{(1 + \beta)}{hie} \text{ expand reactances :-}$$

$$\frac{V_{in}}{I_{in}} = Z_{in} = \left(gm \cdot \frac{1}{j\omega.C_1} \cdot \frac{1}{j\omega.C_2} \right) + \frac{1}{j\omega.C_1} + \frac{1}{j\omega.C_2} \text{ as } j^2 = -1 \text{ then}$$

$$\frac{V_{in}}{I_{in}} = Z_{in} = -gm \cdot \frac{1}{\omega.C_1.C_2} + \frac{1}{j\omega.[C_1.C_2(C_1 + C_2)]}$$

Input impedance (negative) Parallel combination of C1 & C2