

# IC Power Dissipation/Objectives

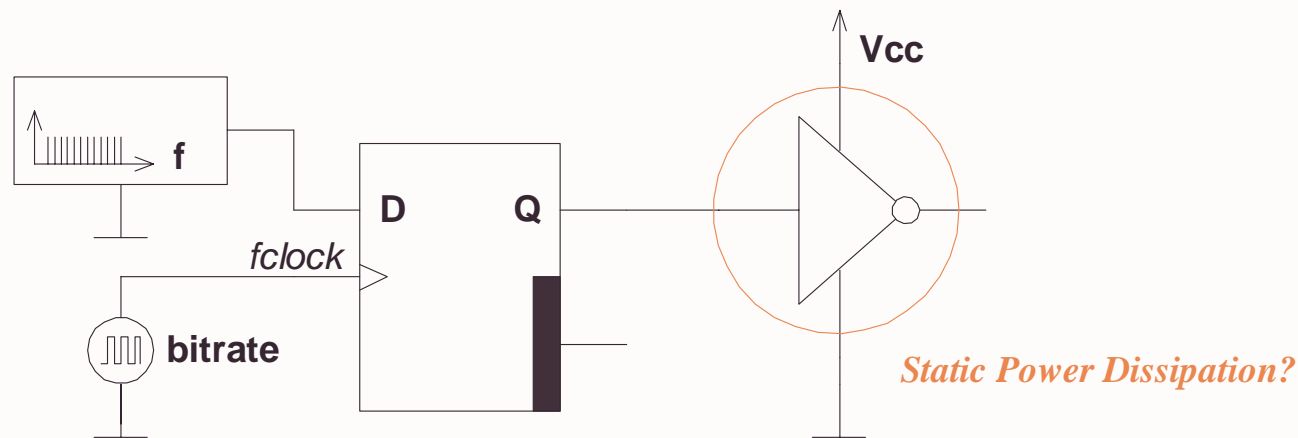
- Power Dissipation in Logic ICs
  - static and dynamic power dissipation
  - input power
  - internal power
  - drive power
  - output power
- Effective cycle frequency for bitstreams

# IC Power Dissipation/General

- Power dissipation in a logic device is only indirectly related to the typical supply current specification on its data sheet ( $P_{cc} = V_{cc} \cdot I_{cc}$ )
- Datasheet specs often ignore additional power dissipation occurring at high speeds or high output loading
- Power dissipation categories of high-speed logic:
  - Input power
  - Internal dissipation
  - Drive circuit dissipation
  - Output power
- Each categorie has two sub-categories
  - static / quiescent (power used to hold a circuit in one logic state or the other)
  - dynamic (power used everytime a logic circuit changes state)

# IC Power Dissipation/Static

- Static power dissipation
  - Power used to hold a circuit in one logic state or the other
  - Depending on situation use
    - » maximum (worst case) power dissipation  
Assumption that IC spends all its time in worst case state
    - » average power dissipation  
Assumption that IC spends equal time in either state
    - » weighted average power dissipation  
Useful if statistical information on states is available...

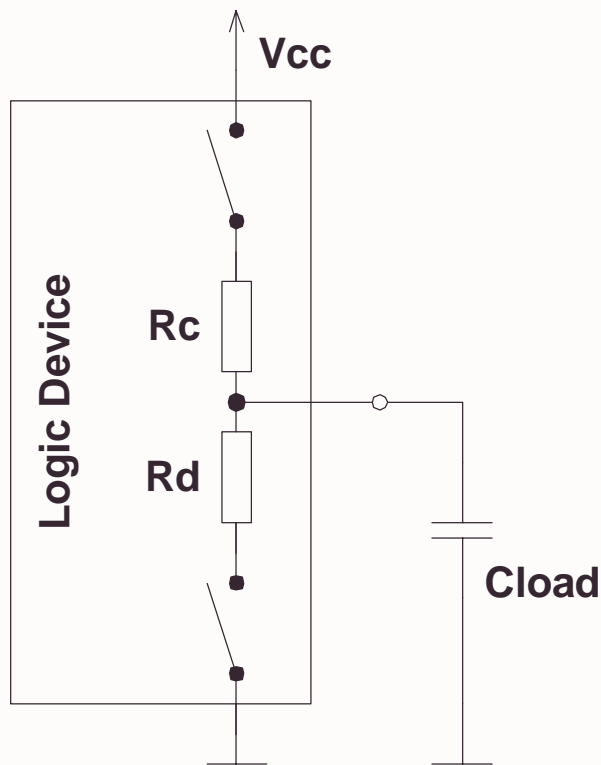


# IC Power Dissipation/Dynamic

Charging/discharging capacitors: Energy budget and power dissipation considerations

Dissipated in  $R_c$  per cycle

$$E_{R_c \text{ charge}} = \frac{1}{2} C \cdot V_{cc}^2$$



Dissipated in  $R_d$  per cycle

$$E_{R_d \text{ discharge}} = \frac{1}{2} C \cdot V_{cc}^2$$

**Note: Regardless of values of  $R_c$  and  $R_d$  !**

Thermal energy in IC per cycle

$$E_{IC \text{ cycle}} = C \cdot V_{cc}^2$$

Average IC Power Dissipation

$$P_{diss} = E_{diss} \cdot f_{cycle} = C \cdot V_{cc}^2 \cdot f_{cycle}$$

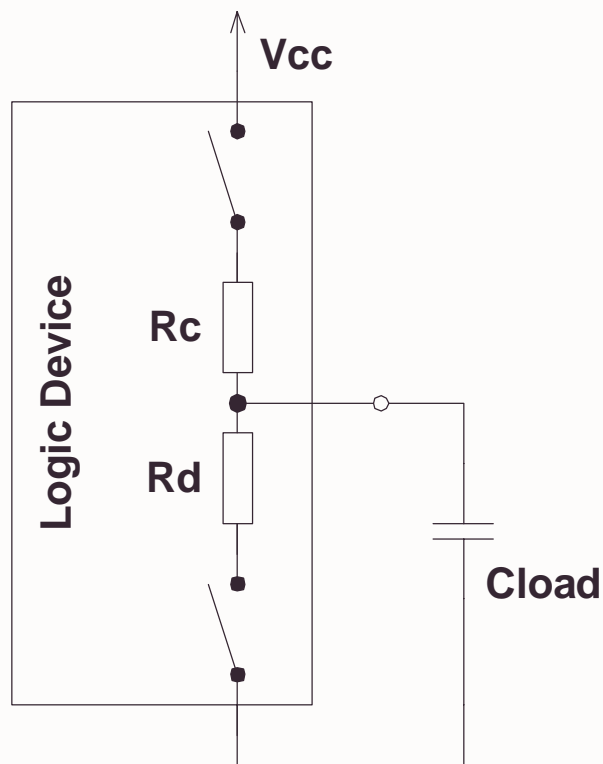
**Equations valid only if**

- Clload fully charged from 0V to  $V_{cc}$
- Clload fully discharged from  $V_{cc}$  to 0V

# IC Power Dissipation/Dynamic

Charging/discharging capacitors: Energy budget and power dissipation considerations

Watch out: Situation changes considerably if load capacitance is pre-charged...

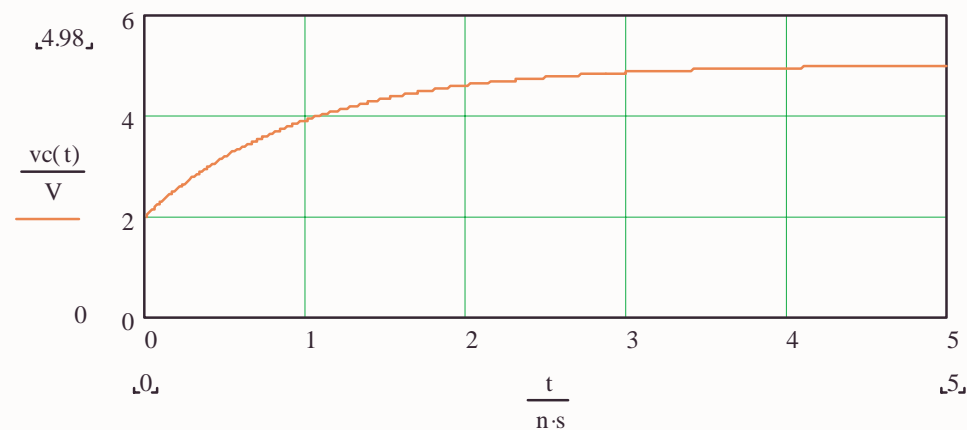


Dissipated in  $R_c$  per charge cycle

$$E_{R_c \text{ charge}} = \frac{1}{2} C \cdot (V_{\text{end}} - V_{\text{start}})^2 = \frac{1}{2} C \cdot \Delta V^2$$

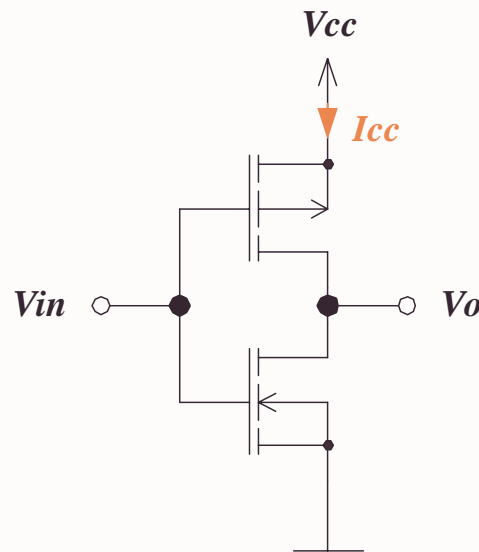
Energy transferred into  $C_{\text{load}}$

$$E_{C_{\text{load}}} = \frac{1}{2} C \cdot (V_{\text{end}}^2 - V_{\text{start}}^2)$$



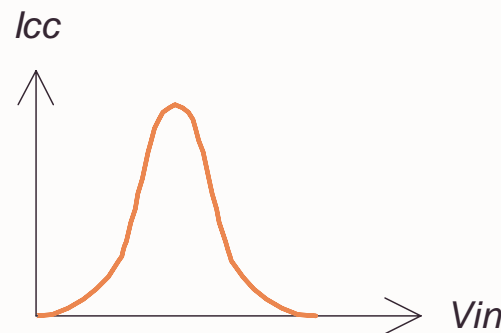
# IC Power Dissipation/Dynamic/Cross-Conduction

- Cross-Conduction common for totem-pole output stages
- During input transitions high-side and low-side switch of totem-pole output conduct simultaneously
- Complex trade-off between speed, output impedance, supply voltage range, etc.
- Particularly pronounced in bipolar TTL (saturated low-side BJT)
- Adverse effect of slow rise and fall times



$$E_{cross-conduction} \propto V_{cc}$$

$$P_{cross-conduction} \propto (f_{cycle}, V_{cc})$$



Modern logic devices featuring very low supply voltages can even suffer from static cross-conduction (significant sub-threshold conduction)

# IC Power Dissipation/Dynamic

- Heat energy due to cross-conduction

$$E_{cross-conduction} \propto V_{CC}$$

- Heat energy due to charge/discharge of circuit capacitances

$$E_{capacitive} \propto V_{CC}^2$$

- Dependency of dynamic power dissipation

$$P_{dynamic} \propto (V_{CC}^{1..2}, f_{cycle})$$

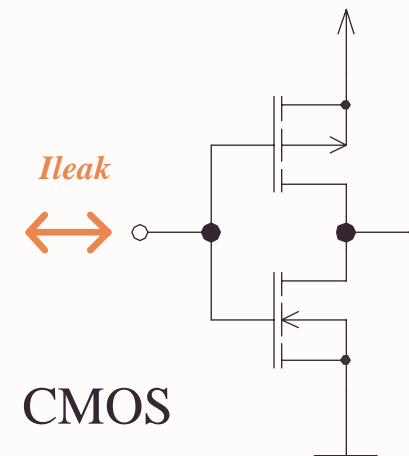
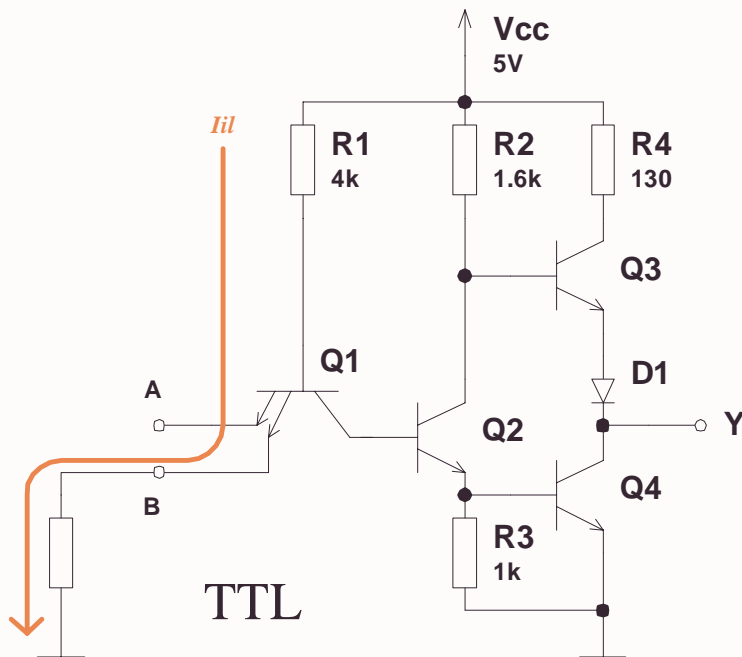
- more typically...

$$P_{dynamic} \propto (V_{CC}^{1.8}, f_{cycle})$$

# IC Power Dissipation/Input Power/Static

- Static Input Power
  - Required to bias and activate the input circuits
  - Often negligible (especially for CMOS)

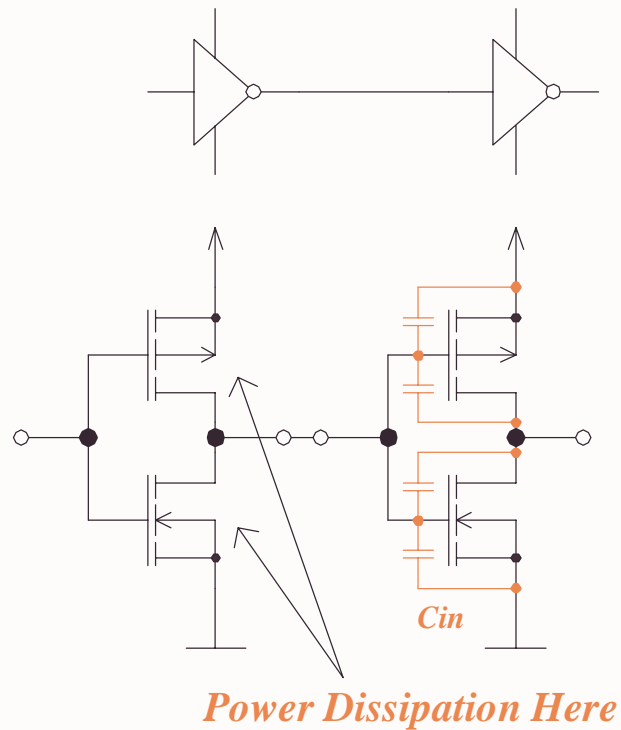
| Logic Family | $I_{IH}$ (max) | $I_{IL}$ (max) |
|--------------|----------------|----------------|
| 74           | 40 $\mu$ A     | -1.6mA         |
| 74LS         | 20 $\mu$ A     | -0.4mA         |
| 74AS         | 20 $\mu$ A     | -0.5mA         |
| 74ALS        | 20 $\mu$ A     | -100 $\mu$ A   |
| 74F          | 20 $\mu$ A     | -0.6mA         |
| 4000         | $\pm 1\mu$ A   | $\pm 1\mu$ A   |
| 74HC/HCT     | $\pm 1\mu$ A   | $\pm 1\mu$ A   |
| 74AC/ACT     | $\pm 1\mu$ A   | $\pm 1\mu$ A   |
| 74AHC/AHCT   | $\pm 1\mu$ A   | $\pm 1\mu$ A   |





# IC Power Dissipation/Input Power/Dynamic

- Dynamic Input Power
  - Due to input capacitance
  - Causes only negligible power dissipation in input stage (i.e. causes power dissipation in output stage of driving circuit...)



| Logic Family | $C_{in}$ (typ) |
|--------------|----------------|
| 74AS         | 3.0pF          |
| ECL          | 3.0pF          |
| 4000         | 5pF            |
| 74HC/HCT     | 3.5pF          |
| 74AC/ACT     | 4.5pF          |
| 74LCX        | 7pF            |

# IC Power Dissipation/Dissipation Constant

- Static and dynamic internal power dissipation
  - Used to bias and switch nodes internal to a logic device (static and dynamic)

Power dissipation constant

$$K_{dynamic} = \frac{P_{internal} - P_{static}}{f_{cycle}} \qquad \{K_{dynamic}\} = W/Hz$$

- Knowing the power dissipation constant, the total internal power dissipation for an arbitrary operating frequency can be approximated
- Does not include extra energy dissipated in the driver caused by a connected load!

$$P_{internal} = P_{static} + K_{dynamic} \cdot f_{cycle}$$

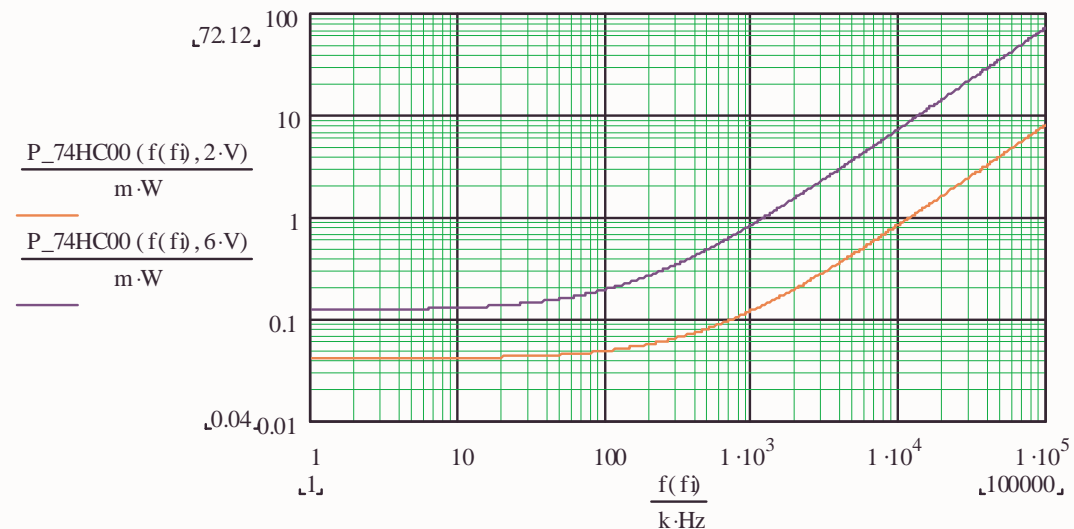
# IC Power Dissipation/Dissipation Constant

- Static and dynamic internal power dissipation
- Some CMOS circuits operate over a wide range of supply voltages. Many datasheets therefore rate their power dissipation in terms of an equivalent capacitance  $C_{PD}$

$$P_{internal} = P_{static} + C_{PD} \cdot V_{CC}^2 \cdot f_{cycle} \quad \{C_{PD}\} = \frac{As}{V}$$

## Example Parameter:

- Fairchild 74HC00
- $I_{CC} = 20\mu A$  (static)
- $V_{CC} = 2V..6V$
- $C_{PD} = 20pF$

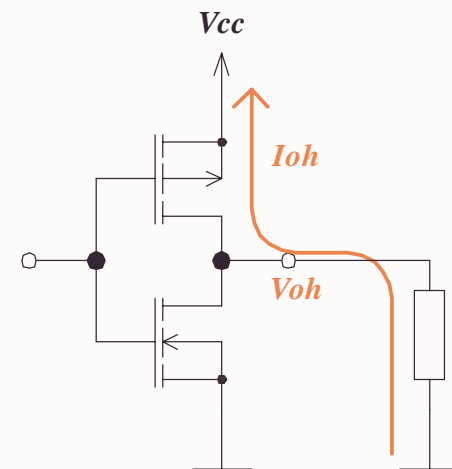
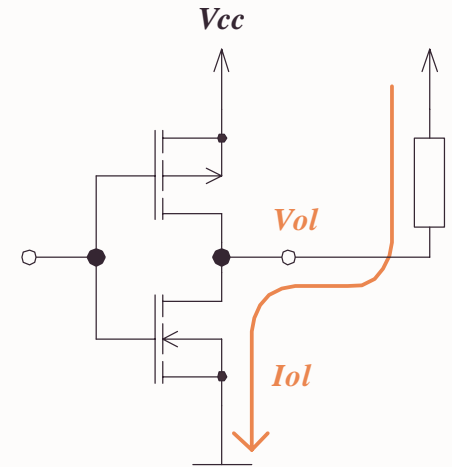


# IC Power Dissipation/Driver Power/Static

- Static driver power dissipation
  - Power dissipation due to load (sink or source) current and the residual voltage across the conducting switch

$$P_{driver_{low}} = V_{ol} \cdot I_{ol}$$

$$P_{driver_{high}} = (V_{cc} - V_{oh}) \cdot |I_{oh}|$$



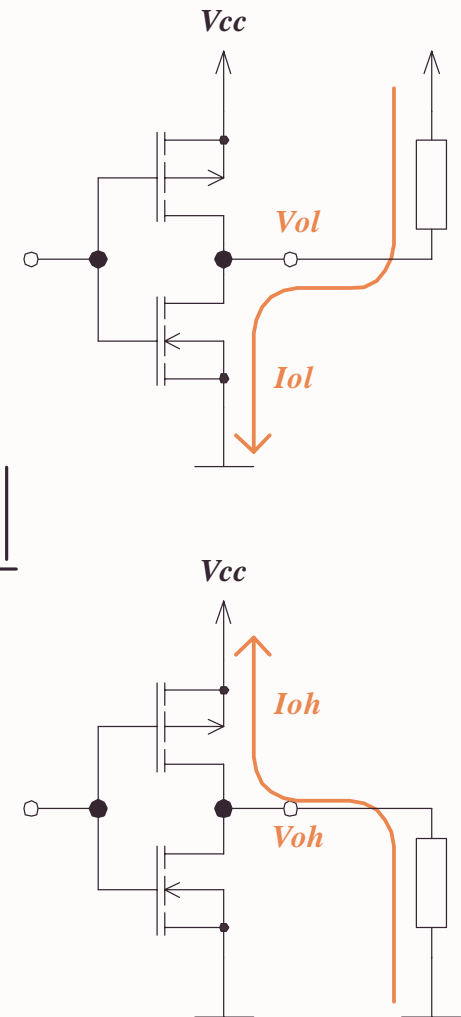
# IC Power Dissipation/Driver Power/Static

- Equivalent driver output impedances and maximum high/low state output currents

| Logic Family | I <sub>OH</sub> (max) | I <sub>OL</sub> (max) |
|--------------|-----------------------|-----------------------|
| 74           | -0.4mA                | 16mA                  |
| 74LS         | -0.4mA                | 8mA                   |
| 74AS         | -2mA                  | 20mA                  |
| 74ALS        | -0.4mA                | 8mA                   |
| 74F          | -1mA                  | 20mA                  |
| 4000         | -0.4mA                | 0.4mA                 |
| 74HC/HCT     | -4mA                  | 4mA                   |
| 74AC/ACT     | -24mA                 | 24mA                  |
| 74AHC/AHCT   | -8mA                  | 8mA                   |

$$R_{driver_{low}} = \frac{|V_{ol}|}{|I_{ol}|}$$

$$R_{driver_{high}} = \frac{|V_{CC} - V_{oh}|}{|I_{oh}|}$$

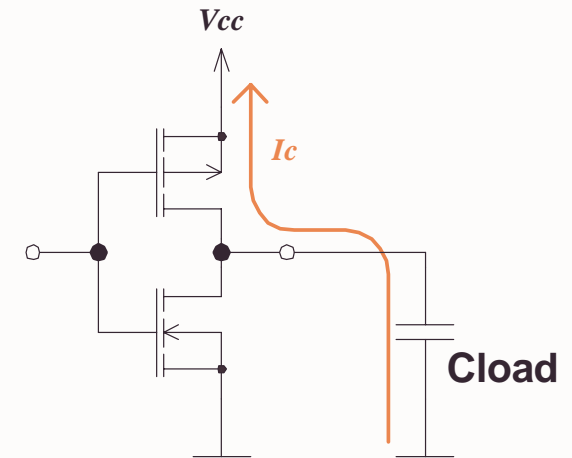
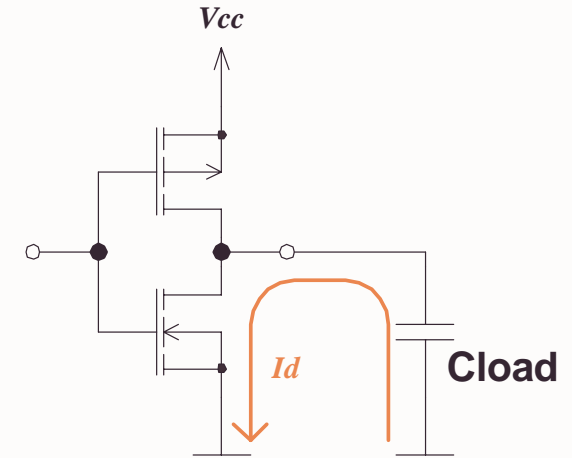


# IC Power Dissipation/Driver Power/Dynamic

- Dynamic driver power dissipation due to
  - charge/discharge of load capacitances

$$P_{driver_{dyn}} = C_{load} \cdot V_{cc}^2 \cdot f_{cycle}$$

$C_{load}$  is typically the sum of input capacitances of gates driven by the output.



# IC Power Dissipation/Output Power

- Output power dissipation
  - mainly in termination resistors

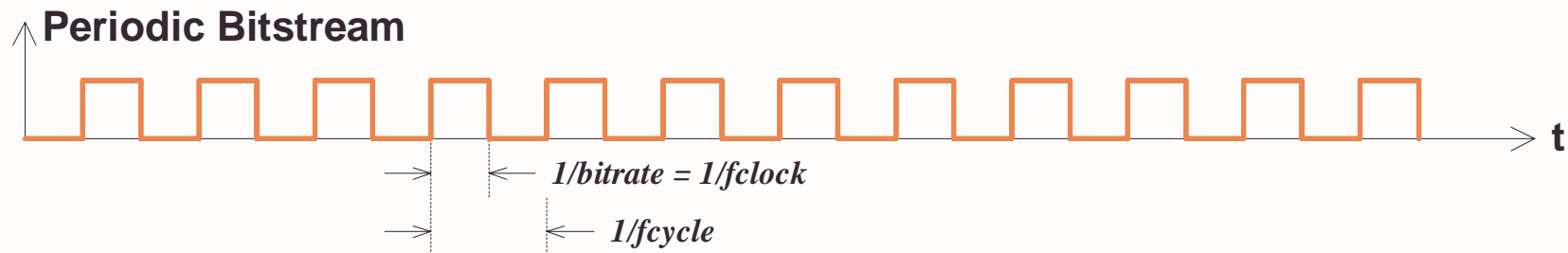
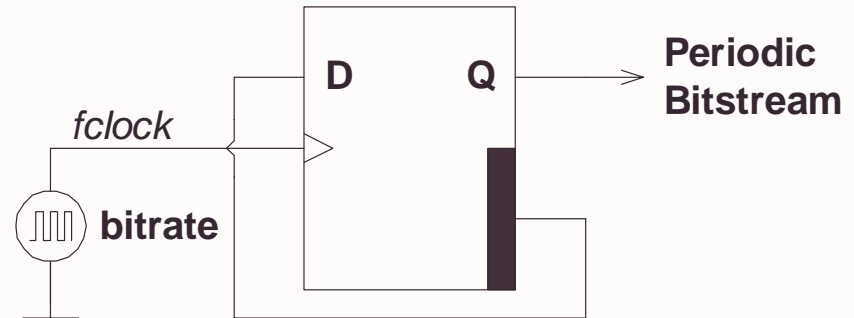


$$P_{R1_{worstcase}} = \frac{V_{cc}^2}{R1} \qquad P_{R1_{average}} = \frac{1}{2} \frac{V_{cc}^2}{R1} \qquad P_{R1_{average}} \neq \frac{\left(\frac{V_{cc}}{2}\right)^2}{R1}$$

**Don't under-estimate power dissipation in termination resistors! Use termination resistor types which can handle the power. Consider worst case conditions for each resistor.**

# Bitstreams/Effective Cycle Frequency/Max

- Worst case cycle frequency of bitstreams
  - Regular pattern
  - One transition per bit



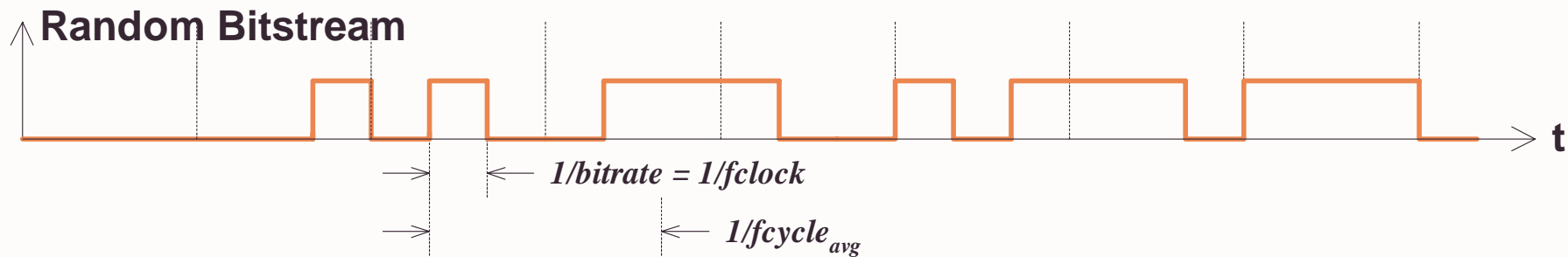
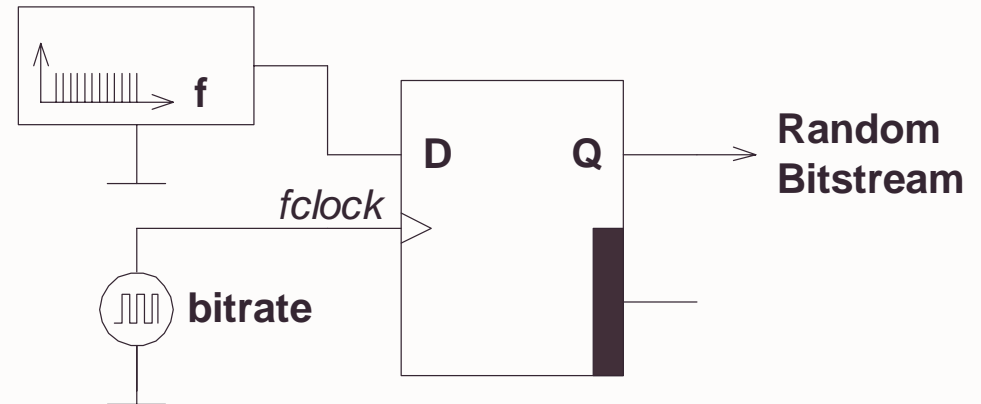
$$f_{cycle} = \frac{bitrate}{2} = \frac{fclock}{2}$$

It takes two transitions (i.e. 2 bits) to complete a cycle !



# Bitstreams/Effective Cycle Frequency/Average

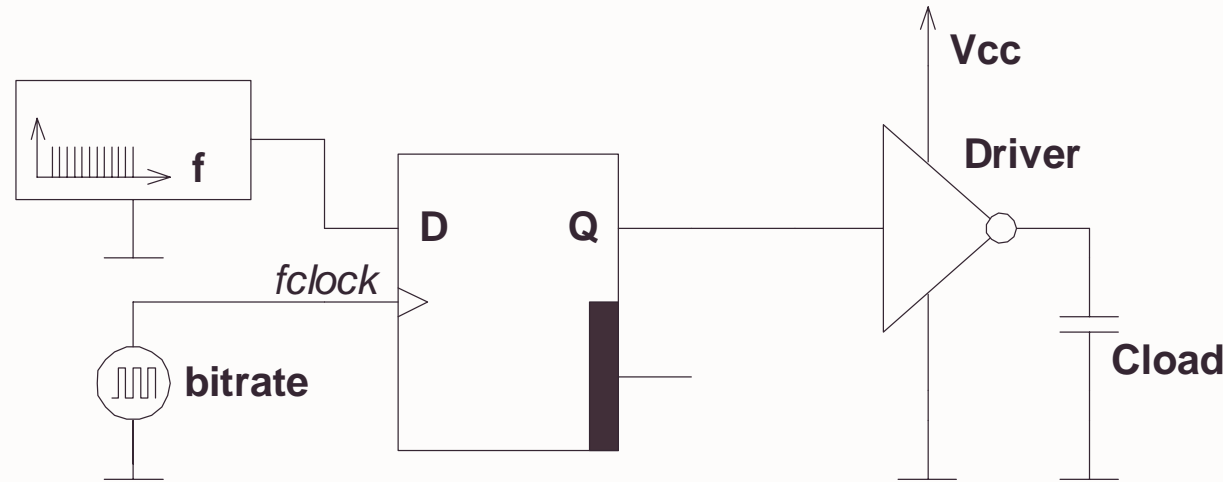
- Average cycle frequency of a random bitstreams
  - Random pattern
  - 0.5 transitions per bit (statistically)



$$f_{\text{cycle}_{\text{avg}}} = \frac{\text{bitrate}}{4} = \frac{f_{\text{clock}}}{4}$$

It takes two transitions to complete a cycle ! On average it will take 4 clock periods to complete a cycle.

# IC Power Dissipation/Driver Power



Worst case and average drive power dissipation:

$$\text{Worst case } P_{diss_{wc}} = \frac{1}{2} Cload \cdot Vcc^2 \cdot bitrate$$

$$\text{Average } P_{diss_{avg}} = \frac{1}{4} Cload \cdot Vcc^2 \cdot bitrate$$

# Thermals/Objectives

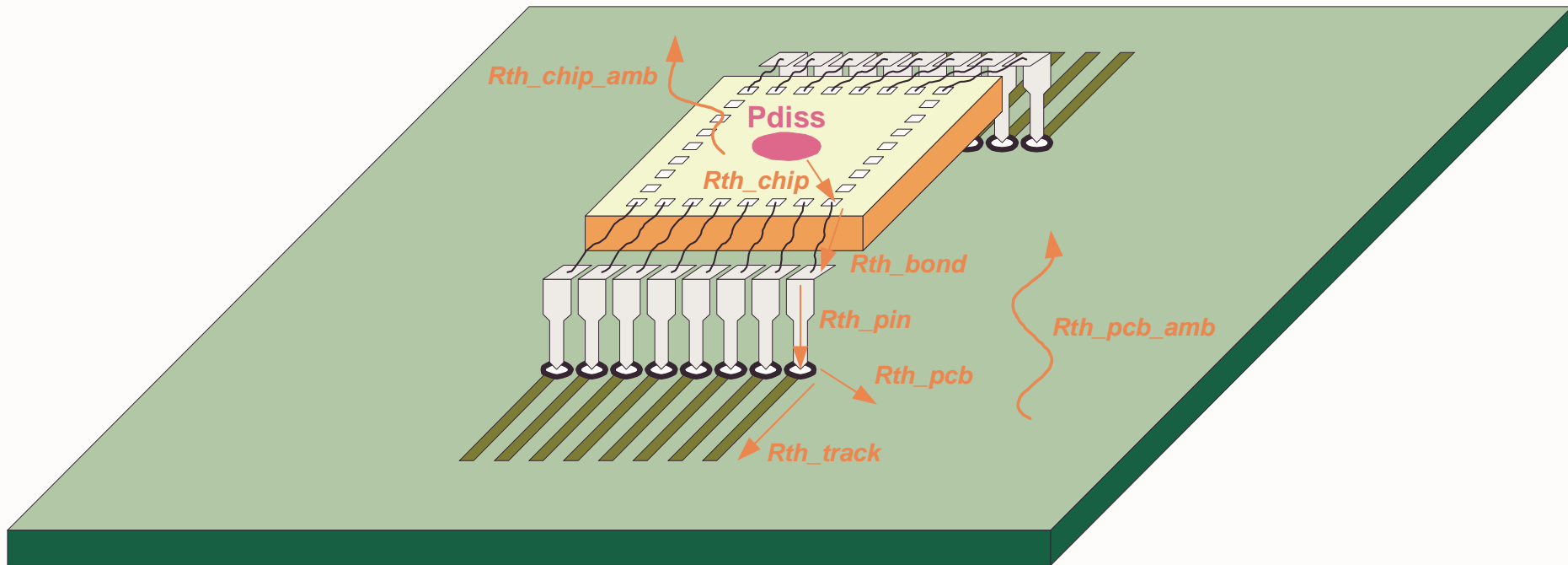
- Thermal Basics
  - Modes of heat transfer
  - Thermal modelling
- IC Packaging
  - Categories (PTH, SMT)
  - Materials
  - Electrical package modelling
- Measurement Techniques
  - Scopes and scope probes

# Thermals/Basics

- Exponential relationship between temperature of a device and its failure rate
- MTBF: Mean Time Between Failures
- Modes of heat transfer
  - Conduction  
Transfer through solid material
  - Convection  
Transfer through a medium of fluid (air for our purposes)
  - Radiation  
Transfer of heat through EM waves

**Arrhenius Law for chemical processes near room temperature:  
Failure rate of devices approx. doubles with every 10°C increase in its temperature...**

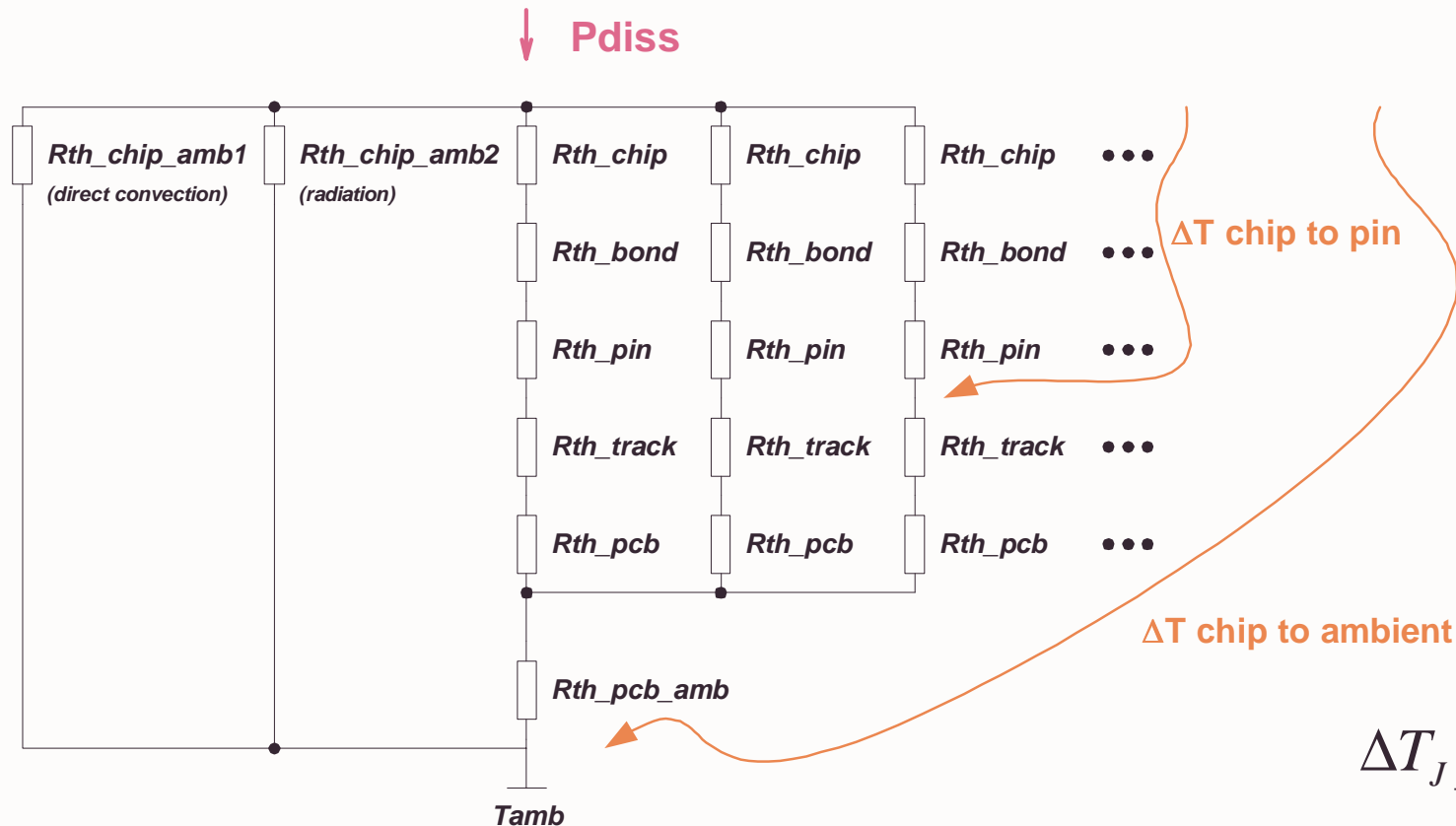
# Thermals/IC Thermal Paths Junction-Ambient



## Thermal arrangement:

Often complex pathes from heat source (e.g. power dissipation on an IC) to heat sink (e.g. ambient).

# Thermals/IC Thermal Paths Junction-Ambient



$$\Delta T_{J\_Pin} = P_{diss} \cdot R_{th_{J\_Pin}}$$

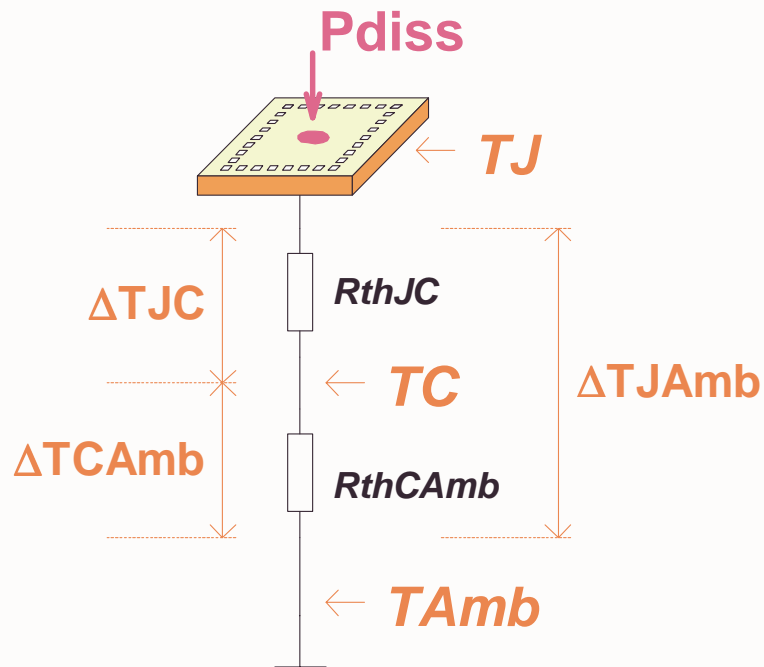
$$\Delta T_{J\_Amb} = P_{diss} \cdot R_{th_{J\_Amb}}$$

## Complex thermal networks

Identification of significant thermal paths is difficult. Quantification of thermal resistances even more so. For proper analysis: Use appropriate SW tools (Thermal FEM Tools, e.g. Flotherm).

# Thermals/IC Thermal Paths Junction-Ambient

- Highly simplified
  - but often adequate for estimating the junction temperature
  - Thermal resistances  $R_{thJC}$  and/or  $R_{thJAmb}$  specified in most IC datasheets



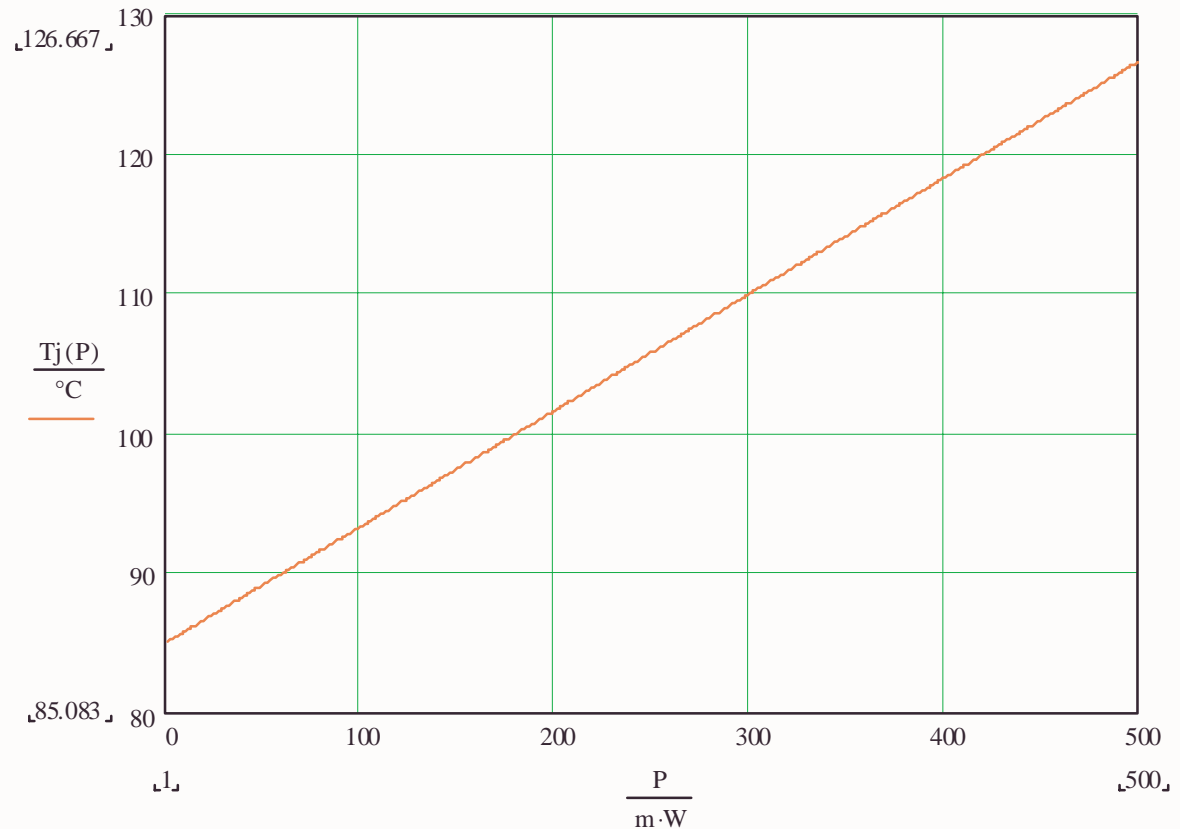
$$T_J = T_C + P_{diss} \cdot R_{thJC}$$

$$T_J = T_{amb} + P_{diss} \cdot (R_{thJC} + R_{thCAmb})$$

# Thermals/IC Thermal Paths Junction-Ambient

## Example Parameter:

- Fairchild 74HC00
- $R_{thJA}=83.3\text{K/W}$  (still air)
- $T_{ambmax}=85^\circ\text{C}$
- $P_{dmax}=500\text{mW}$  (SO14)



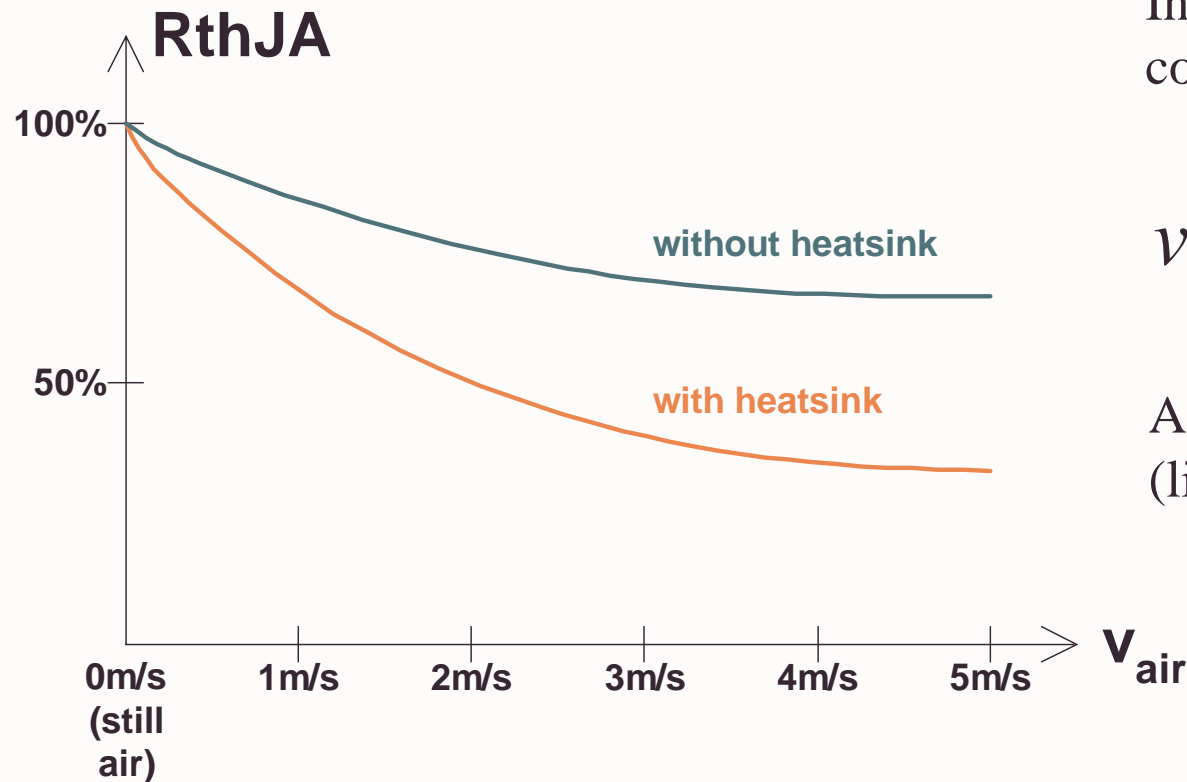
$$T_J = T_{amb} + P_{diss} \cdot (R_{th_{JC}} + R_{th_{CAmb}})$$



# Thermals/IC Thermal Paths Junction-Ambient

## Forced Air Convection:

Typical behavior thermal resistance versus air velocity in forced air convection:



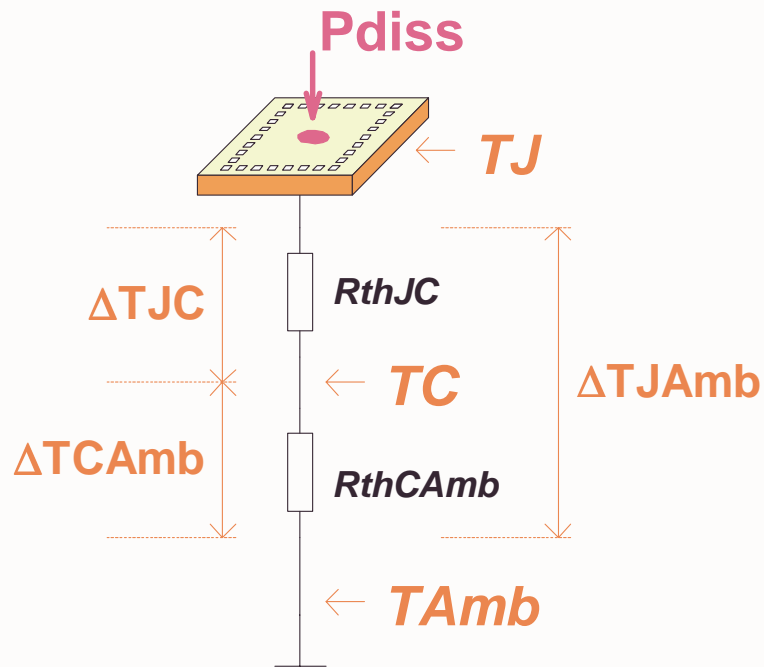
In a typical system with forced air convection:

$$v_{air} = 0.5 \frac{m}{s} \text{ .. } 2 \frac{m}{s}$$

Air velocity often quoted in LFPM (linear feet per minute)...

$$100 \text{ LFPM} \approx 0.5 \frac{m}{s}$$

# Thermals/IC Thermal Paths Junction-Ambient



- Watch Out!

- Definition of ambient temperature  $T_{Amb}$  can be ambiguous (ambient temperature of a system vs local ambient of an electronic sub-assembly)
- IC manufacturers tend to specify  $R_{thJAmb}$ . Read the small-print!  $R_{thJAmb}$  is often specified using unrealistically large copper areas around the IC
- Always: Verify your thermal predictions through thermal measurements!
  - » IR camera
  - » no-contact Laser thermometer
  - » thermo-couples